# Electroluminescent Lamp Drivers 

## EL Lamp Applications



- Thermostats
- Weight Scales
- Cellular Phones
- Digital Compasses
- HPCs (Handheld PCs)


## Pocket Organizer



Pen Computer

- Temperature Monitors
- Automotive Dashboards
- GPS Handheld Receivers
- PDAs (Personal Digital Assistants)
- Watches and Alarm Clocks
- Test and Medical Equipment
- TV/VCR/Audio/Cable Box Remote Controllers


## IMP, Inc. - Company Profile

IMP, Inc. designs, manufacturers and markets standard-setting analog integrated circuits and specialty analog wafer foundry processes for data communications interface and power management applications in computer, communications and control systems. IMP products are sold through a worldwide network of sales representatives and distributors.

## Company Facilities

IMP headquarters and ISO 9001 certified wafer fabrication and test facility are located in San Jose, California. Product development centers are in Pleasanton, California, and Lee, New Hampshire. Sales offices are in San Jose, California; Dallas, Texas; West Kingston, Rhode Island and Singapore.

## Principal Markets

Data Communications Interface - Internal system data communications circuits, including single-ended (SE), low voltage differential (LVD) and multimode (SE/LVD) Small Computer Systems Interface (SCSI) terminators.

Power Management - Circuits that generate, distribute, protect and manage the thermal and power consumption characteristics of hand-held, portable and battery-powered systems. Portable computers, mobile and wireless communication devices and battery-powered medical systems are typical market segments. Example products include electroluminescent lamp drivers, microprocessor supervisors, voltage monitors, low dropout voltage regulators, and high-frequency switching converters.

## Wafer Fabrication and Manufacturing Services

High-volume, analog and mixed-signal wafer foundry services on low-power, high-voltage ( 100 V and above), CMOS, BiCMOS, and EEPROM processes, including turnkey packaging and test capabilities. Fabrication services include database production using IMP standard processes, and porting of customerowned technology.

## For More Information

Visit the IMP web site at www.impweb.com; email info@impinc.com or contact IMP headquarters at 408.432.9100/800-438-3722.

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## IMP Electroluminescent Lamp Drivers

IMP electroluminescent lamp drivers incorporate four EL lamp driving functions on-chip. These are the boost switch-mode power supply, its high-frequency oscillator, the high-voltage H-bridge lamp driver and its low-frequency oscillator. Few external components are needed: one inductor, one diode, one capacitor and two resistors. The resistors allow independent adjustment of boost converter frequency and EL lamp drive frequency. Adjustable lamp drive frequency allows control over lamp color and power dissipation. All devices can be disabled for power saving.

All devices are available in chip form and small MicroSO and SO packages. Tape and reel shipment is available without additional cost.


## IMP525: Single Cell Battery Powered Electroluminescent Lamp Driver/ Inverter

The IMP525 Electroluminescent (EL) lamp driver is designed for systems that must operate down to 1 V and below. The input supply voltage range is 0.9 V to 2.5 V . Typical output lamp drive voltage is 112 V peak-to-peak. EL lamps of up to 6 nF capacitance can be driven to high brightness.

A disable mode puts the chip into a low current-drain state. When disabled, quiescent current drops to $1 \mu \mathrm{~A}$ typical with a $\mathrm{V}_{\mathrm{DD}}$ of 1.5 V . Connecting $\mathrm{R}_{\mathrm{SW}}$, the oscillator frequency setting resistor, to ground, can disable the chip. A disable pad, accessible only on the die, can also be used to disable the driver (active low). An internal circuit shuts down the switching regulator when the lamp drive voltage exceeds 112 V peak-to-peak. This conserves power and extends battery life.

## Key Features

- Wide operating voltage range - from 0.9 V to 2.5 V
- Simple design requires few passive components
- 112V peak-to-peak typical AC output voltage
- Adjustable output frequency controls lamp color and power consumption
- Adjustable converter frequency minimizes circuit power consumption
- Disable mode extends battery life
- Disable current $1 \mu \mathrm{~A}$ typical
- Compact MicroSO package and die option
- Same pinout as IMP803 EL Driver Product Line Summary


## IMP560: Power Efficient EL Lamp Driver

The IMP560 is designed for systems with modest EL lamp drive voltage requirements. It is ideal for low ambient light applications or where small lamps are used. With just one-half the inductor current of the IMP803, the IMP560 reduces system power consumption and extends battery life. Input supply voltage range is 2.0 V to 6.5 V and quiescent current is a low $420 \mu \mathrm{~A}$. Typical EL lamp drive voltage is 120 V peak-to-peak.

An internal circuit shuts down the switching regulator when the lamp drive voltage exceeds 120 V peak-to-peak. This conserves power and extends battery life.
A disable mode puts the chip into a low current drain mode. With a 3.0 V supply, quiescent current drops to 200 nA maximum, 50 nA typical.

## Key Features

- 120V peak-to-peak typical AC output voltage
- Low input current (w/ inductor current)...... 12 mA
- Low disabled input current..... 50 nA
- Wide operating voltage range - from 2.0 V to 6.5 V
- Simple design requires few passive components
- Adjustable output lamp frequency controls lamp color and power consumption
- Adjustable converter frequency for minimum power consumption
- IMP803 pin compatible
- MicroSO package option


## IMP803: High-Voltage EL Lamp Driver

The IMP803 drives EL lamps of up to 30 nF capacitance to high brightness. EL lamps with capacitance greater than 30 nF can be driven but will be less bright. The typical regulated output voltage that is applied to the EL lamp is 180 V peak-to-peak.
The IMP803 operates over a 2.0 V to 6.5 V supply voltage range. A regulated, low-power source can supply the low quiescent current of the IMP803. The inductor may be driven from an independent, unregulated supply voltage in dual supply applications. An internal circuit shuts down the switching regulator when the lamp drive voltage reaches 180 V peak-to-peak. This conserves power and extends battery life.

## Key Features

- Low Power: $420 \mu \mathrm{~A}$ typical $\mathrm{V}_{\text {DD }}$ current
- Wide operating voltage range - from 2.0 V to 6.5 V
- 180V peak-to-peak typical AC output voltage
- Large output load capability - drive lamps with more than 30nF capacitance
- Adjustable output lamp frequency for control of lamp color, lamp life, and power consumption
- Adjustable converter frequency to minimize power consumption
- Device can be Enabled/ Disabled
- Low quiescent current - 20nA (disabled)
- High-Voltage CMOS Process
- MicroSO package option


## EL Lamp Driver Product Summary Table

| Part | Input Voltage <br> Range (V) | Packages | Low Power <br> Disable Mode | Typical Output <br> Voltage (VP) | Adjustable Lamp <br> Drive and Boost <br> Frequency | Regulated <br> Output Voltage |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IMP525 | 0.9 to 2.5 | MicroSO \& SO | Yes | 112 | Yes | Yes |
| IMP527 | 0.9 to 2.5 | MicroSO \& SO | Yes | 180 | Yes | Yes |
| IMP528 | 2 to 6.5 | MicroSO \& SO | Yes | 220 | Yes | Yes |
| IMP560 | 2 to 6.5 | MicroSO \& SO | Yes | 120 | Yes | Yes |
| IMP803 | 2 to 6.5 | MicroSO \& SO | Yes | 180 | Yes | Yes |

[^0]ELD/B_t06

EL Driver Product Line Summary

## EL Lamp Driver Development Kits

Several demonstration boards and evaluation kits are available to reduce time-to-market. The kits are available by calling IMP Customer Service at 408.432.9100.

| Item | Device/Package | Description |
| :---: | :---: | :--- |
| IMP803EV1 | IMP803LG | Evaluation board. Has all components plus battery and lamp. |
| IMPxxxDBM | Any MicroSO | Development board. For evaluating IC sample(s) in-circuit. |
| IMPxxxDBS | Any SO | Development board. For evaluating IC sample(s) in-circuit. |
| IMPELD003 | Any SO | Evaluation board with pin jacks for part changes. No R, C, L or battery. |

Note: "xxx" denotes any driver; 525, 527,528, 560 or 803.
ELD/B_105

## Electroluminescent Lamp Applications and Benefits

Liquid Crystal Displays (LCDs) must be lighted for viewing in darkness or low ambient light conditions. Typically, light is projected forward from the back of the LCD display. EL lamps are popular backlights for liquid crystal displays and keypads because EL lamps are flexible, lightweight, thin, vibration and impact resistant, and can be shaped into small, complex or irregular forms. EL lamps evenly light an area without creating "bright-spots".

Since EL lamps typically consume much less current than incandescent bulbs or light emitting diodes (LEDs), their low power consumption, low heat generation and flexibility make them ideal for battery powered portable applications.

EL lamp backlighting applications include: keyless entry systems; audio/video equipment remote controllers; PDA keyboards and displays; timepieces and watches; LCD displays in cellular phones, pagers, and handheld Global Positioning Systems (GPS); face illumination for instrumentation; assistance lighting for buildings; and decorative lighting for sign-displays and merchandising displays.

## Typical EL Lamp Applications

- PDAs
- Safety illumination
- Portable instrumentation
- Battery-operated displays
- LCD modules
- Toys
- Automotive displays
- Cellular phones
- Night lights
- Audio and TV remote control units
- Panel meters
- Pagers


## EL Driver Product Updates

New product information and application notes can be obtained by visiting the IMP web site at www.impweb.com or by sending email to info@impinc.com.

## Single Cell Battery Powered Electroluminescent Lamp Driver/Inverter

The IMP525 is an Electroluminescent (EL) lamp driver designed for systems that must operate down to 1 volt and below. The input supply voltage range is 0.9 V to 2.5 V . Typical output lamp drive voltage is 112 V . All four EL lamp-driving functions are on-chip. These are the switch-mode power supply, its high-frequency oscillator, the high-voltage H-bridge lamp driver and its low-frequency oscillator. EL lamps of up to 6 nF capacitance can be driven to high brightness.

The circuit requires few external components; one inductor, one diode, one capacitor and two resistors. The resistors set the frequency for the two oscillators.

A disable mode puts the chip into a low current-drain state. When disabled, quiescent current drops to $1 \mu \mathrm{~A}$ typical with a $\mathrm{V}_{\mathrm{DD}}$ of 1.5 V . The chip can be disabled by connecting $\mathrm{R}_{\mathrm{SW}}$, the oscillator frequency setting resistor, to ground. A disable pad (active low), accessible only on the die, can also be used to disable the driver.

An internal circuit shuts down the switching regulator when the lamp drive voltage exceeds 112 V peak-to-peak. This conserves power and extends battery life.

The IMP525 is available in MicroSO and SO-8 packages and in die form.

## Key Features

- Wide operating voltage range - from 0.9 V to 2.5 V
- Simple design requires few passive components
- 112V peak-to-peak typical AC output voltage
- Adjustable output frequency controls lamp color and power consumption
- Adjustable converter frequency minimizes circuit power consumption
- Disable mode extends battery life
- Disable current $1 \mu \mathrm{~A}$ typical
- Compact MicroSO package option


## Applications

- Audio/ TV remote control units
- Pagers/ Cellular phones
- PDAs
- Clocks and radios
- Portable GPS receivers
- LCD modules
- Toys

Block Diagram


## Pin Configuration

## SO/MicroSO



## Ordering Information

| Part Number | Input Voltage | Regulated Output Voltage | Temperature Range | Pins-Package |
| :--- | :---: | :---: | :---: | :---: |
| IMP525EMA | 0.9 V to 2.5 V | YES | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 -MicroSO |
| IMP525ESA | 0.9 V to 2.5 V | YES | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 -SO |
| IMP525/D | 0.9 V to 2.5 V | YES | $25^{\circ} \mathrm{C}$ | Dice |

Add /T to ordering part number for Tape and Reel.

## Absolute Maximum Ratings

Supply Voltage, $\mathrm{V}_{\mathrm{DD}}, \mathrm{V}_{\text {RSw-OSC }}$ and $\mathrm{V}_{\text {REL-OSC }} \ldots-0.5 \mathrm{~V}$ to +3.5 V

Power Dissipation (SO package) . . . . . . . . . . . . 400mW
Power Dissipation (MicroSO package) . . . . . . . 300mW

Note: All voltages are referenced to GND.
These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability.

## Electrical Characteristics

Unless otherwise noted, $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}, \mathrm{R}_{\mathrm{SW}}=1 \mathrm{M} \Omega, \mathrm{R}_{\mathrm{EL}}=1.0 \mathrm{M} \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ON-resistance of MOS Switch | $\mathrm{R}_{\mathrm{DS} \text { (ON) }}$ | $\mathrm{I}=50 \mathrm{~mA}$ |  |  | 15 | $\Omega$ |
| Operating Voltage |  |  | 0.9 |  | 2.5 | V |
| Output Voltage at $\mathrm{C}_{\mathrm{S}}$ | $\mathrm{V}_{\text {CS }}$ | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$, See Figure 1, Table 1 | 52 | 58 | 65 | V |
| Output Voltage at $\mathrm{C}_{\mathrm{S}}$ | $V_{\text {CS }}$ | $\mathrm{V}_{\mathrm{DD}}=0.9 \mathrm{~V}$, See Figure 1, Table 2 |  | 50 |  | V |
| Output Voltage Peak-to-Peak | $\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}$ | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$, See Figure 1 | 104 | 112 | 124 | $V_{\text {P-P }}$ |
| Quiescent $V_{D D}$ Supply Current, Disabled (Disable pin available on die only) | $\mathrm{I}_{\text {QDIS }}$ | Disable $=$ HIGH |  | 70 |  | nA |
| Quiescent V ${ }_{\text {DD }}$ Supply Current, Disabled | $\mathrm{I}_{\text {QDIS }}$ | $\begin{aligned} & \mathrm{R}_{\mathrm{SW}-\mathrm{OSC}}=\mathrm{GND} \\ & \mathrm{~V}_{\mathrm{DD}}=1.5 \mathrm{~V} \end{aligned}$ |  | 1.0 | 2.0 | $\mu \mathrm{A}$ |
| Input Current at $\mathrm{V}_{\mathrm{DD}}$ Pin | $\mathrm{I}_{\mathrm{DD}}$ | $\mathrm{V}_{\mathrm{DD}}=0.9 \mathrm{~V}$ to 1.5 V |  |  | 1.5 | mA |
| Input Current: IDD ${ }_{\text {Plus }}$ Inductor Current | 1 IN | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$ |  | 23 | 32 | mA |
| $\mathrm{V}_{\mathrm{A}-\mathrm{B}}$ Output Drive Frequency | $\mathrm{f}_{\mathrm{EL}}$ | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$, See Figure 1, Table 1 |  | 500 |  | Hz |
| Boost Converter Switching Frequency | $\mathrm{f}_{\text {Sw }}$ | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$, See Figure 1, Table 1 |  | 26 |  | kHz |
| Switching Duty Cycle | $\mathrm{D}_{\text {SW }}$ | $V_{D D}=1.5 \mathrm{~V}$, See Figure 1 |  | 87.5 |  | \% |
| Disable Input LOW Voltage (Disable pin available on die only) | $V_{\text {DISL }}$ |  | GND |  | 0.2 | V |
| Disable Input HIGH Voltage (Disable pin available on die only) | $\mathrm{V}_{\text {DISH }}$ |  | $\mathrm{V}_{\mathrm{DD}}-0.5 \mathrm{~V}$ |  | $V_{D D}$ | V |

IMP525

## Typical Characteristics




Boost Converter Switching Period


## Pin Descriptions

| Pin Number | Name | Function |
| :---: | :---: | :---: |
| 1 | $V_{D D}$ | Positive voltage supply for the IMP525. Inductor L may be connected here or to a separate supply. |
| 2 | $\mathrm{R}_{\text {SW-OSC }}$ | Switch-mode resistor pin. Switching frequency is determined by external resistor $\mathrm{R}_{\mathrm{SW}}$, connected between pin 2 and $\mathrm{V}_{\mathrm{DD}}$. |
| 3 | Cs | Boost converter storage capacitor. The voltage across the EL lamp is equal to twice the voltage at $\mathrm{C}_{\mathrm{s}}$. |
| 4 | Lx | Connection to flyback inductance, L. |
| 5 | GND | Ground pin. |
| 6 | $V_{B}$ | EL lamp drive. The lamp is connected to a high-voltage bridge circuit with $\mathrm{V}_{\mathrm{B}}$ providing the complementary connection to $\mathrm{V}_{\mathrm{A}}$. |
| 7 | $V_{\text {A }}$ | EL lamp drive. (See above) |
| 8 | $\mathrm{R}_{\text {EL-OSC }}$ | The EL lamp oscillator frequency-setting pin. The frequency is controlled by resistor $\mathrm{R}_{\mathrm{EL}}$, connected from pin 8 to $V_{D D}$. |
| Disable Pad | DIS | Available only in die form. Setting DIS HIGH disables the chip. |

## External Components

| External Component | Description and Selection Guide |
| :---: | :---: |
| Diode | A fast reverse recovery diode, with BV $>100$, such as a 1 N4148. |
| Capacitor $\mathrm{C}_{\mathrm{s}}$ | The high voltage capacitor that stores the inductive energy transferred through the catch diode. A 100 volt capacitor between 10 nF and 100 nF is recommended. |
| Resistor $\mathrm{R}_{\mathrm{EL}}$ | The EL lamp oscillator frequency-setting resistor. $\mathrm{R}_{\mathrm{EL}}$ is connected between pin 8 and $\mathrm{V}_{\mathrm{DD}}$, providing a frequency inversely proportional to $\mathrm{R}_{\mathrm{EL}}$; as $\mathrm{R}_{\mathrm{EL}}$ increases, the EL lamp frequency decreases along with the current drawn by the lamp. Lamp color is also determined by this frequency. A $1 \mathrm{M} \Omega$ resistor between the $\mathrm{R}_{\mathrm{EL} \text {-osc }}$ pin and the $\mathrm{V}_{\mathrm{DD}}$ supply results in a lamp frequency around 500 Hz . |
| Resistor Rsw | Switching Oscillator frequency-setting resistor. $R_{\text {Sw }}$ is connected between the $R_{\text {Sw-osc }}$ pin and the $V_{D D}$ supply. The switching frequency is inversely proportional to the resistor value, dropping as the resistance increases. |
| Capacitor $\mathrm{C}_{\text {sw }}$ | This is an optional noise-suppression capacitor connected from ground to the Rsw-osc pin. A 100 pF capacitor is recommended. |
| Inductor L | The inductor provides the voltage boost needed by means of inductive "flyback". The internal MOSFET switch alternately opens and closes the ground connection for the inductor at the $\mathrm{L}_{\mathrm{x}}$ pin. When the switch opens, the inductor potential will forward-bias the diode and the current will pass through to the storage capacitor $\mathrm{C}_{\mathrm{s}}$, charging it to a high voltage. <br> As the value of the inductor is increased, the switching frequency set by $\mathrm{R}_{\mathrm{SW}}$ should also be increased to prevent saturation. In general, smaller value inductors that can handle more current are more desirable when larger-area EL lamps must be driven. <br> A small electrolytic capacitor ( $10 \mu \mathrm{~F}, 16 \mathrm{~V}$ ), normally present across the inductor supply $\mathrm{V}_{\mathbb{N}}$, will likely eliminate the need for $\mathrm{C}_{s w}$. |

## Application Information

## Test Circuit

Figure 1 shows the IMP525 configured to drive an EL lamp,
represented as a 3 nF capacitor.


Figure 1. Test Circuit

Table 1. $\mathrm{V}_{\mathrm{IN}}=1.5 \mathrm{~V}$

| Component | Connections | Value | Description |
| :---: | :---: | :---: | :--- |
| $R_{\mathrm{SW}}$ | $\mathrm{V}_{\mathrm{DD}}, \mathrm{R}_{\mathrm{SW} \text {-OSC }}$ | $1 \mathrm{M} \Omega$ | Boost converter oscillator bias resistor |
| $\mathrm{R}_{\mathrm{EL}}$ | $\mathrm{V}_{\mathrm{DD}}, \mathrm{R}_{\mathrm{EL}-\mathrm{OSC}}$ | $1 \mathrm{M} \Omega$ | EL lamp driver oscillator bias resistor |
| L | $\mathrm{V}_{\mathrm{DD}}, \mathrm{Lx}^{2}$ | $330 \mu \mathrm{H}^{2}$ | Boost converter inductor |
| $\mathrm{C}_{\mathrm{S}}$ | $\mathrm{C}_{\mathrm{S}}, \mathrm{GND}$ | $0.1 \mu \mathrm{~F} / 100 \mathrm{~V}$ | Boost converter storage capacitor |
| D | $\mathrm{L}_{\mathrm{x}}, \mathrm{C}_{\mathrm{S}}$ | 1 N 4148 | Switching diode |
| $\mathrm{C}_{\mathrm{SW}}$ | $\mathrm{R}_{\mathrm{SW}-\mathrm{OSC}}, \mathrm{GND}$ | 0.1 nF | Noise-suppression capacitor |

Notes. 2. Murata LQH4N331K04 (8.2 2 max. DCR)

Table 2. $\mathrm{V}_{\mathrm{IN}}=0.9 \mathrm{~V}$

| Component | Connections | Value | Description |
| :---: | :---: | :---: | :--- |
| $R_{S W}$ | $\mathrm{~V}_{\mathrm{DD}}, \mathrm{R}_{\mathrm{SW}-\mathrm{OSC}}$ | $1.0 \mathrm{M} \Omega$ | Boost converter oscillator bias resistor |
| $\mathrm{R}_{\mathrm{EL}}$ | $\mathrm{V}_{\mathrm{DD}}, \mathrm{R}_{\mathrm{EL}-\mathrm{OSC}}$ | $2.62 \mathrm{M} \Omega$ | EL lamp driver oscillator bias resistor |
| L | $\mathrm{V}_{\mathrm{DD}}, \mathrm{L}^{3}$ | $680 \mu \mathrm{H}^{3}$ | Boost converter inductor |
| $\mathrm{C}_{S}$ | $\mathrm{C}_{\mathrm{S}}, G N D$ | $0.1 \mu \mathrm{~F} / 100 \mathrm{~V}$ | Boost converter storage capacitor |
| D | $\mathrm{L}_{\mathrm{x}}, \mathrm{C}_{\mathrm{S}}$ | 1 N 4148 | Switching diode |
| $\mathrm{C}_{S W}$ | $\mathrm{R}_{\mathrm{SW}-\mathrm{OSC}}, G N D$ | 0.1 nF | Noise-suppression capacitor |

Notes. 3. Coilcraft DS1608C-684 (2.2 $\Omega$ max. DCR)

## Enable/ Disable Operation

Figure 2 shows how the IMP525 can be enabled via a logic gate that connects $\mathrm{R}_{\mathrm{SW}}$ to $\mathrm{V}_{\mathrm{DD}}$, and disabled by connecting it to ground.

The IMP525 can also be disabled using a pad on the die. The Disable function pin is not available in packaged parts.

| Enable/Disable Table |  |
| :---: | :---: |
| $\mathbf{R}_{\text {SW }}$ Connection | IMP525 State |
| $\mathrm{V}_{\mathrm{DD}}$ | Enabled |
| Ground | Disabled |


| Disable PAD Connection <br> (Available only with dice) | IMP525 State |
| :---: | :---: |
| HIGH $\left(\mathrm{V}_{\mathrm{DD}}\right)$ | Disabled |
| LOW $(\mathrm{Ground})$ | Enabled |


cmos



* Optional

Figure 2. Enable/Disable Operation

## High Voltages Present

The IMP525 generates high voltages and caution should be exercised.

## Inductor Manufacturers

| Manufacturer | Series | USA Phone Number |
| :--- | :--- | :---: |
| Toko | D52FU | $(847)$ 297-0070 |
| Coilcraft | DS1608, DO1608, DT1608 | $(847)$ 639-6400 |
| River Electronics | FLC32 | $(310)$ 320-7488 |
| Murata | LQH4N | $(800)$ 831-9172 |

## Single Cell Battery Powered Electroluminescent Lamp Driver/Inverter

The IMP527 is an Electroluminescent (EL) lamp driver designed for systems that must operate down to 1 volt and below. The input supply voltage range is 0.9 V to 2.5 V . Typical output lamp drive voltage is 180 V . All four EL lamp-driving functions are on-chip. These are the switchmode power supply, its high-frequency oscillator, the high-voltage H -bridge lamp driver and its low-frequency oscillator. EL lamps of up to 6 nF capacitance can be driven to high brightness.
The circuit requires few external components; one inductor, one diode, one capacitor and two resistors. The resistors set the frequency for the two oscillators.

A disable mode puts the chip into a low current-drain state. When disabled, quiescent current drops to $1 \mu \mathrm{~A}$ typical with a $\mathrm{V}_{\mathrm{DD}}$ of 1.5 V . The chip can be disabled by connecting $R_{S W}$, the oscillator frequency setting resistor, to ground. A disable pad (active low), accessible only on the die, can also be used to disable the driver.

An internal circuit shuts down the switching regulator when the lamp drive voltage exceeds 180 V peak-to-peak. This conserves power and extends battery life.
The IMP527 is available in MicroSO and SO-8 packages and in die form.
Block Diagram

## Key Features

- Wide operating voltage range - from 0.9 V to 2.5 V
- Simple design requires few passive components
- 180V peak-to-peak typical AC output voltage
- Adjustable output frequency controls lamp color and power consumption
- Adjustable converter frequency minimizes circuit power consumption
- Disable mode extends battery life
- Disable current 1 1 A typical
- Compact MicroSO package option


## Applications

- Audio/ TV remote control units
- Pagers/ Cellular phones
- PDAs
- Clocks and radios
- Portable GPS receivers
- LCD modules
- Toys


Pin Configuration


## Ordering Information

| Part Number | Input Voltage | Regulated Output Voltage | Temperature Range | Pins-Package |
| :--- | :---: | :---: | :---: | :---: |
| IMP527EMA | 0.9 V to 2.5 V | YES | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $8-\mathrm{MicroSO}$ |
| IMP527ESA | 0.9 V to 2.5 V | YES | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $8-\mathrm{SO}$ |
| IMP527/D | 0.9 V to 2.5 V | YES | $25^{\circ} \mathrm{C}$ | Dice |

## Absolute Maximum Ratings

Supply Voltage, $\mathrm{V}_{\mathrm{DD}}, \mathrm{V}_{\text {RSw-OSC }}$ and $\mathrm{V}_{\text {REL-OSC }} \ldots-0.5 \mathrm{~V}$ to +3.5 V
Storage Temperature Range $\qquad$ $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Power Dissipation (SO package) 400 mW
Power Dissipation (MicroSO package) $\qquad$ 300 mW

Note: All voltages are referenced to GND.

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability.

## Electrical Characteristics

Unless otherwise noted, $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}, \mathrm{R}_{\mathrm{SW}}=1 \mathrm{M} \Omega, \mathrm{R}_{\mathrm{EL}}=1 \mathrm{M} \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ON-resistance of MOS Switch | $\mathrm{R}_{\mathrm{DS} \text { (ON) }}$ | $\mathrm{I}=50 \mathrm{~mA}$ |  |  | 15 | $\Omega$ |
| Operating Voltage |  |  | 0.9 |  | 2.5 | V |
| Output Voltage at $\mathrm{C}_{\mathrm{S}}$ | $\mathrm{V}_{\text {CS }}$ | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$, See Figure 1, Table 1 | 80 | 90 |  | V |
| Output Voltage at $\mathrm{C}_{\mathrm{S}}$ | $\mathrm{V}_{\text {CS }}$ | $V_{D D}=0.9 \mathrm{~V}$, See Figure 1, Table 2 |  | 50 |  | V |
| Output Voltage Peak-to-Peak | $V_{A}-V_{B}$ | $V_{D D}=1.5 \mathrm{~V}$, See Figure 1 |  | 180 |  | $\mathrm{V}_{\text {P-P }}$ |
| Quiescent $\mathrm{V}_{\mathrm{DD}}$ Supply Current, Disabled (Disable pin available on die only) | $\mathrm{I}_{\text {QDIS }}$ | Disable $=\mathrm{HIGH}$ |  | 70 |  | nA |
| Quiescent V ${ }_{\text {DD }}$ Supply Current, Disabled | $\mathrm{I}_{\text {QDIS }}$ | $\begin{aligned} & \mathrm{R}_{\mathrm{SW}-\mathrm{OSC}}=\mathrm{GND} \\ & \mathrm{~V}_{\mathrm{DD}}=1.5 \mathrm{~V} \end{aligned}$ |  | 1.0 | 2.0 | $\mu \mathrm{A}$ |
| Input Current at $\mathrm{V}_{\text {DD }}$ Pin | $\mathrm{I}_{\mathrm{DD}}$ | $\mathrm{V}_{\mathrm{DD}}=0.9 \mathrm{~V}$ to 1.5 V |  |  | 1.5 | mA |
| Input Current: IDD ${ }_{\text {Plus }}$ Inductor Current | 1 IN | $V_{D D}=1.5 \mathrm{~V}$, See Figure 1, Table 1 |  | 26 | 32 | mA |
| $\mathrm{V}_{\text {A-B }}$ Output Drive Frequency | $\mathrm{f}_{\mathrm{EL}}$ | $V_{D D}=1.5 \mathrm{~V}$, See Figure 1, Table 1 |  | 500 |  | Hz |
| Boost Converter Switching Frequency | $\mathrm{f}_{\text {Sw }}$ | $V_{D D}=1.5 \mathrm{~V}$, See Figure 1, Table 1 |  | 26 |  | kHz |
| Switching Duty Cycle | $\mathrm{D}_{\text {SW }}$ | $\mathrm{V}_{\mathrm{DD}}=1.5 \mathrm{~V}$, See Figure 1 |  | 87.5 |  | \% |
| Disable Input LOW Voltage (Disable pin available on die only) | $\mathrm{V}_{\text {DISL }}$ |  | GND |  | 0.2 | V |
| Disable Input HIGH Voltage (Disable pin available on die only) | $\mathrm{V}_{\text {DISH }}$ |  | $\mathrm{V}_{\mathrm{DD}}-0.5 \mathrm{~V}$ |  | $V_{D D}$ | V |

## Typical Characteristics

EL Lamp Drive Frequency


EL Lamp Drive Period


Boost Converter Switching Frequency


Boost Converter Switching Period


## Pin Descriptions

| Pin Number | Name | Function |
| :---: | :---: | :---: |
| 1 | $V_{D D}$ | Positive voltage supply for the IMP527. Inductor L may be connected here or to a separate supply. |
| 2 | Rsw-osc | Switch-mode resistor pin. Switching frequency is determined by external resistor $\mathrm{R}_{\mathrm{sw}}$, connected between pin 2 and $\mathrm{V}_{\mathrm{DD}}$. |
| 3 | Cs | Boost converter storage capacitor. The voltage across the EL lamp is equal to twice the voltage at $\mathrm{C}_{\mathrm{s}}$. |
| 4 | Lx | Connection to flyback inductance, L. |
| 5 | GND | Ground pin. |
| 6 | $V_{B}$ | EL lamp drive. The lamp is connected to a high-voltage bridge circuit with $\mathrm{V}_{\mathrm{B}}$ providing the complementary connection to $\mathrm{V}_{\mathrm{A}}$. |
| 7 | $V_{\text {A }}$ | EL lamp drive. (See above) |
| 8 | $\mathrm{R}_{\text {EL-osc }}$ | The EL lamp oscillator frequency-setting pin. The frequency is controlled by resistor $\mathrm{R}_{\mathrm{EL}}$, connected from pin 8 to $V_{D D}$. |
| Disable Pad | DIS | Available only in die form. Setting DIS HIGH disables the chip. |

## External Components

| External Component | Description and Selection Guide |
| :---: | :---: |
| Diode | A fast reverse recovery diode, with BV > 100, such as a 1 N4148. |
| Capacitor Cs | The high voltage capacitor that stores the inductive energy transferred through the catch diode. A 100 volt capacitor between 10 nF and 100 nF is recommended. |
| Resistor $\mathrm{R}_{\mathrm{EL}}$ | The EL lamp oscillator frequency-setting resistor. $\mathrm{R}_{\mathrm{EL}}$ is connected between pin 8 and $\mathrm{V}_{\mathrm{DD}}$, providing a frequency inversely proportional to $\mathrm{R}_{\mathrm{EL}}$; as $\mathrm{R}_{\mathrm{EL}}$ increases, the EL lamp frequency decreases along with the current drawn by the lamp. Lamp color is also determined by this frequency. A $1 \mathrm{M} \Omega$ resistor between the $\mathrm{R}_{\mathrm{EL} \text {-osc }}$ pin and the $\mathrm{V}_{\mathrm{DD}}$ supply results in a lamp frequency around 500 Hz . |
| Resistor Rsw | Switching Oscillator frequency-setting resistor. $R_{\text {sw }}$ is connected between the $R_{\text {Sw-osc }}$ pin and the $V_{D D}$ supply. The switching frequency is inversely proportional to the resistor value, dropping as the resistance increases. |
| Capacitor $\mathrm{C}_{\text {sw }}$ | This is an optional noise-suppression capacitor connected from ground to the $\mathrm{R}_{\text {sw-osc }}$ pin. A 100 pF capacitor is recommended. |
| Inductor L | The inductor provides the voltage boost needed by means of inductive "flyback". The internal MOSFET switch alternately opens and closes the ground connection for the inductor at the $L_{x}$ pin. When the switch opens, the inductor potential will forward-bias the diode and the current will pass through to the storage capacitor $\mathrm{C}_{\mathrm{s}}$, charging it to a high voltage. <br> As the value of the inductor is increased, the switching frequency set by $R_{S w}$ should also be increased to prevent saturation. In general, smaller value inductors that can handle more current are more desirable when larger-area EL lamps must be driven. <br> A small electrolytic capacitor ( $10 \mu \mathrm{~F}, 16 \mathrm{~V}$ ), normally present across the inductor supply $\mathrm{V}_{\mathbb{N}}$, will likely eliminate the need for $\mathrm{C}_{S W}$. |

## Application Information

## Test Circuit

Figure 1 shows the IMP527 configured to drive an EL lamp, represented as a 3 nF capacitor.


Figure 1. Test Circuit
Table 1. $\mathrm{V}_{\mathrm{IN}}=1.5 \mathrm{~V}$

| Component | Connections | Value | Description |
| :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\text {SW }}$ | $\mathrm{V}_{\text {DD }}, \mathrm{R}_{\text {SW-OSC }}$ | $1 \mathrm{M} \Omega$ | Boost converter oscillator bias resistor |
| $\mathrm{R}_{\mathrm{EL}}$ | $\mathrm{V}_{\mathrm{DD}}, \mathrm{R}_{\mathrm{EL}-\mathrm{OSC}}$ | $1 \mathrm{M} \Omega$ | EL lamp driver oscillator bias resistor |
| L | $\mathrm{V}_{\mathrm{DD}}, L_{\text {c }}{ }^{2}$ | $330 \mu \mathrm{H}^{2}$ | Boost converter inductor |
| Cs | $\mathrm{C}_{\mathrm{s}}$, GND | $0.1 \mu \mathrm{~F} / 100 \mathrm{~V}$ | Boost converter storage capacitor |
| D | Lx, Cs | 1N4148 | Switching diode |
| $\mathrm{C}_{\text {SW }}$ | $\mathrm{R}_{\text {SW-osc }}$, GND | 0.1 nF | Noise-suppression capacitor (optional) |

Notes. 2. Murata LQH4N331K04 (8.2 2 max. DCR)

Table 2. $\mathrm{V}_{\mathrm{IN}}=0.9 \mathrm{~V}$

| Component | Connections | Value | Description |
| :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\text {Sw }}$ | $\mathrm{V}_{\text {DD }}, \mathrm{R}_{\text {SW-OSc }}$ | $1 \mathrm{M} \Omega$ | Boost converter oscillator bias resistor |
| $\mathrm{R}_{\text {EL }}$ | $\mathrm{V}_{\text {DD }}, \mathrm{R}_{\text {EL-OSC }}$ | $2.62 \mathrm{M} \Omega$ | EL lamp driver oscillator bias resistor |
| L | $V_{\text {DD }}, L_{\text {x }}{ }^{3}$ | $680 \mu \mathrm{H}^{3}$ | Boost converter inductor |
| Cs | $\mathrm{C}_{\mathrm{s}}$, GND | $0.1 \mu \mathrm{~F} / 100 \mathrm{~V}$ | Boost converter storage capacitor |
| D | $\mathrm{L}_{\mathrm{x}}, \mathrm{C}_{\text {S }}$ | 1N4148 | Switching diode |
| $\mathrm{C}_{\text {sw }}$ | $\mathrm{R}_{\text {Sw-osc }}$, GND | 0.1 nF | Noise-suppression capacitor (optional) |

Notes. 3. Coilcraft DS1608C-684 (2.2ת max. DCR)

## Enable/ Disable Operation

Figure 2 shows how the IMP527 can be enabled via a logic gate that connects $\mathrm{R}_{\mathrm{SW}}$ to $\mathrm{V}_{\mathrm{DD}}$, and disabled by connecting it to ground.

The IMP527 can also be disabled using a pad on the die. The Disable function pin is not available in packaged parts.

| Enable/Disable Table |  |
| :---: | :---: |
| R $_{\text {SW }}$ Connection | IMP527 State |
| $\mathrm{V}_{\text {DD }}$ | Enabled |
| Ground | Disabled |


| Disable PAD Connection <br> (Available only with dice) | IMP527 State |
| :---: | :---: |
| HIGH (VD) | Disabled |
| LOW (Ground) | Enabled |



* Optional

Figure 2. Enable/Disable Operation

## High Voltages Present

The IMP527 generates high voltages and caution should be exercised.

## Inductor Manufacturers

| Manufacturer | Series | USA Phone Number |
| :--- | :--- | :---: |
| Toko | D52FU | $(847)$ 297-0070 |
| Coilcraft | DS1608, DO1608, DT1608 | $(847)$ 639-6400 |
| River Electronics | FLC32 | $(310)$ 320-7488 |
| Murata | LQH4N | $(800)$ 831-9172 |

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Block Diagram

## Power Management

## High-Voltage EL Lamp Driver

## - 220 V $_{\text {PP }}$ Drive

The IMP528 is an Electroluminescent (EL) lamp driver with the four EL lamp driving functions on-chip. These are the switch-mode power supply, its high-frequency oscillator, the high-voltage H-bridge lamp driver and its low-frequency oscillator. The IMP528 drives EL lamps of up to 50nF capacitance to high brightness; EL lamps with capacitances greater than 50 nF can be driven, but will be lower in light output. The typical regulated output voltage that is applied to the EL lamp is 220 V peak-topeak. The circuit requires few external components; a single inductor, single diode, two capacitors and three resistors. Two of these resistors set the frequency for two internal oscillators.

Unlike other EL lamp drivers, the IMP528 does not require an external protection resistor in series with the EL lamp.

The IMP528 operates over a 2.0 V to 6.5 V supply voltage range. A regulated, low-power source can supply the low quiescent current of the IMP528. The inductor may be driven from an independent, unregulated supply voltage in dual supply applications.
An internal circuit shuts down the switching regulator when the lamp drive voltage reaches 220 V peak-to-peak. This conserves power and extends battery life.

The IMP528 is available in MicroSO and SO-8 packages and in die or wafer form. the

## Key Features

220V peak-to-peak typical AC output voltage

- Low Power: $420 \mu \mathrm{~A}$ typical $\mathrm{V}_{\mathrm{DD}}$ current
- Wide operating voltage range-from 2.0 V to 6.5 V
- Large output load capability - drives lamps with more than 50 nF capacitance
- Eliminates external protection resistor in series with EL lamp
- Adjustable output lamp frequency for control of lamp color, lamp life, and power consumption
- Adjustable converter frequency to minimize power consumption
- High-Voltage CMOS Process
- MicroSO package option


## Applications

- GPS units/ Pagers/ Cellular phones
- PDAs/ Handheld computers
- Safety illumination
- Portable instrumentation
- Battery-operated displays
- LCD modules
- Toys



## Pin Configuration

## SO/ MicroSO



## Ordering Information

| Part Number | Input Voltage | Regulated Output Voltage | Temperature Range | Pins-Package |
| :--- | :---: | :---: | :---: | :---: |
| IMP528ESA | 2.0 V to 6.5 V | Yes | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $8-\mathrm{SO}$ |
| IMP528EMA | 2.0 V to 6.5 V | Yes | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $8-\mathrm{MicroSO}$ |
| IMP528/D | 2.0 V to 6.5 V | Yes | $25^{\circ} \mathrm{C}$ | Dice |
| IMP528/D1 | 2.0 V to 6.5 V | Yes | $25^{\circ} \mathrm{C}$ | Dice |

Add /T to ordering part number for Tape and Reel.

## Absolute Maximum Ratings

| $\mathrm{V}_{\text {DD }}$ | -0.5 V to +7.0 V |
| :---: | :---: |
| $\mathrm{V}_{\text {RSW-OSC }}$ and $\mathrm{V}_{\text {ReL-OSC }}$ | -0.5 V to $\mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V}$ |
| $\mathrm{V}_{\mathrm{CS}}, \mathrm{L}_{\mathrm{X}}$ | -0.5 V to +120 V |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Power Dissipation (SO) | 400 mW |
| Power Dissipation (MicroSO) | 300 mW |
| $\mathrm{V}_{\mathrm{A}}, \mathrm{V}_{\mathrm{B}}$ | -0.5 V to $\mathrm{V}_{\mathrm{CS}}(\mathrm{pin} 3)$ |

$V_{D D}$ . -0.5 V to $\mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V}$
$\mathrm{V}_{\mathrm{CS}}$, $\mathrm{L}_{\mathrm{X}}$
-0.5 V to +120 V
Operating Temperature Range
$-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Power Dissipation (SO)
400 mW
$V_{A}, V_{B}$
-0.5 V to $\mathrm{V}_{\mathrm{CS}}(\operatorname{pin} 3)$

## Note: All voltages are referenced to GND.

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability.

## Electrical Characteristics

Unless otherwise noted, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{R}_{\mathrm{SW}}=910 \mathrm{k} \Omega, \mathrm{R}_{\mathrm{EL}}=2.7 \mathrm{M} \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ON-resistance of MOS Switch | R ${ }_{\text {DS(ON) }}$ | $\mathrm{I}=100 \mathrm{~mA}$ |  | 3.0 | 8 | $\Omega$ |
| Output Voltage Regulation | $\mathrm{V}_{\text {cs }}$ | $\mathrm{V}_{\mathrm{DD}}=2.0$ to 6.5 V |  | 110 |  | V |
| Output Voltage Peak-to-peak (in regulation) | $V_{A}-V_{B}$ | $V_{D D}=2.0$ to 6.5 V |  | 220 |  | V |
| Input Current at $\mathrm{V}_{D D}$ Pin | $\mathrm{I}_{\mathrm{DD}}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 420 | 700 | $\mu \mathrm{A}$ |
| Input Current at $\mathrm{V}_{D D}$ Pin | IDD | $\mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V}$ |  | 500 | 750 | $\mu \mathrm{A}$ |
| Quiescent $\mathrm{V}_{\mathrm{DD}}$ Supply Current, Disabled | $I_{\text {DD }}$ | $V_{\text {RSW }- \text { osc }}<100 \mathrm{mV}$ |  | 20 | 200 | nA |
| Input Current: I IDD Plus Inductor Current | $\mathrm{I}_{\mathrm{N}}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 21 | 31 | mA |
| Output Voltage at $\mathrm{V}_{\text {CS }}$ | $\mathrm{V}_{\mathrm{cs}}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 110 |  | V |
| $\mathrm{V}_{\text {A-B }}$ Output Drive Frequency | $\mathrm{f}_{\text {EL }}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 250 |  | Hz |
| Switching Frequency | $\mathrm{f}_{\text {Sw }}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 61 |  | kHz |
| Switching Duty Cycle | $\mathrm{D}_{\text {sw }}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 88 |  | \% |

## Pin Descriptions

| Pin Number | Name | Function |
| :---: | :---: | :---: |
| 1 | $\mathrm{V}_{\mathrm{DD}}$ | Positive voltage supply for the IMP528. Inductor L may be connected here or to a separate unregulated supply. |
| 2 | $\mathrm{R}_{\text {SW-OSC }}$ | Switch-mode resistor pin. Switching frequency is determined by an external resistor, $\mathrm{R}_{\mathrm{Sw}}$. |
| 3 | $\mathrm{C}_{\mathrm{S}}$ | Boost converter storage capacitor. The voltage across the EL lamp is equal to twice the voltage at $\mathrm{C}_{\mathrm{s}}$. |
| 4 | $L_{x}$ | Connection to flyback inductance, L. |
| 5 | GND | Ground pin. |
| 6 | $V_{B}$ | EL lamp drive. The lamp is connected in a high-voltage bridge circuit with $\mathrm{V}_{\mathrm{B}}$ providing the complementary connection to $\mathrm{V}_{\mathrm{A}}$. The peak-to-peak AC voltage across the EL lamp is thus two times $\mathrm{V}_{\text {cs }}$. |
| 7 | $\mathrm{V}_{\mathrm{A}}$ | EL lamp drive. (See above) |
| 8 | $\mathrm{R}_{\text {EL-OSC }}$ | The EL lamp oscillator frequency setting pin. The oscillator frequency is controlled by external resistor REL. |

## External Components

| External Component | Description and Selection Guide |
| :---: | :---: |
| Diode | Catch diode. A fast reverse recovery diode, with BV > 150V, such as an FDLL400 (150V). |
| Capacitor $\mathrm{C}_{\text {s }}$ | This is the high voltage capacitor that stores the inductive energy transferred through the catch diode. A capacitor with WV $>120 \mathrm{~V}$ between 10 nF and 100 nF is recommended. |
| Resistor REL | The EL lamp oscillator frequency setting resistor. This resistor, connected between the $R_{E L-O S C}$ pin and $V_{D D}$, provides an oscillator frequency inversely proportional to $R_{E L}$; as $R_{E L}$ increases, the EL lamp frequency decreases along with the current drawn by the lamp. Lamp color is also determined by this frequency. A $2.7 \mathrm{M} \Omega$ resistor between the $\mathrm{R}_{\mathrm{EL} \text {-osc }}$ pin and the $V_{D D}$ supply results in a lamp frequency around 250 Hz . |
| Resistor R ${ }_{\text {Sw }}$ | Switching Oscillator frequency setting resistor. The switching oscillator resistor is connected between the $R_{\text {Sw-OSC }}$ pin and the $\mathrm{V}_{\mathrm{DD}}$ supply. The switching frequency is inversely proportional to the resistor value, dropping as the resistance increases. |
| Inductor L | The inductor provides the voltage boost needed by means of inductive "flyback". The internal MOSFET switch alternately opens and closes the ground connection for the inductor at the $L_{x}$ pin. When this internal switch opens, the inductor potential will forward-bias the catch diode and the current will pass through the storage capacitor $\mathrm{C}_{\mathrm{s}}$, charging it to a high voltage. <br> Smaller inductors are preferred to prevent saturation. As the value of the inductor increases (and the series DC resistance of the inductor decreases), the switching frequency set by $\mathrm{R}_{\mathrm{SW}}$ should be increased to prevent saturation. In general, smaller value inductors that can handle more current are more desirable when larger area EL lamps must be driven. |

## Application Information

## Test and Application Circuit, 3.0V

Figure 1 shows the IMP528 configured to drive an EL lamp with a 3.0 V input.


Figure 1. 3.0V Application

## Dual Supply Operation with 1.5 V Battery

The IMP528 can also be operate from a single battery cell when a regulated voltage higher than 2.0 V is also available. This dual supply configuration, shown in Figure 2, uses the regulated voltage to operate the IMP528 while the energy for the highvoltage boost circuit comes from the battery.


Figure 2. Dual Supply Operation

## Switch Resistance

The IMP528 inductor switch resistance is typically below $3.5 \Omega$, as shown in Figure 3.

## High-Voltages Present



The IMP528 generates high voltages and caution should be exercised.

Figure 3. Boost Switch ON-Resistance

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## Power Management

A disable mode puts the chip into a low current drain mode. With a 3.0V supply, quiescent current drops to 200 nA maximum, 50 nA typical. The chip is disabled by connecting the oscillator frequency setting resistor $\mathrm{R}_{\text {SW }}$ to ground.

The IMP560 is available in MicroSO and SO-8 packages and in die or wafer form.

## Power Efficient EL Lamp Driver

The IMP560 is an Electroluminescent (EL) lamp driver designed for systems with low EL lamp drive voltage requirements. It is ideal for low tems with low EL lamp drive voltage requirements. It is ideal for low
ambient light applications or where small lamps are used. With just onehalf the inductor current of the IMP803, the IMP560 reduces system power consumption and extends battery life. Input supply voltage range is 2.0 V to 6.5 V and quiescent current is a low $420 \mu \mathrm{~A}$. Typical EL lamp drive voltage is $\pm 56 \mathrm{~V}$.

All four EL lamp-driving functions are on-chip. These are the switchmode power supply, its high-frequency oscillator, the high-voltage H-bridge lamp driver and its low-frequency oscillator. EL lamps of up to 6 nF capacitance can be driven to high brightness.

The circuit requires few external components; a single inductor, a single diode, two capacitors and three resistors. Two of these resistors set the frequencies for two internal oscillators. An internal circuit shuts down the switching regulator when the lamp drive voltage exceeds 120 V peak-to-peak. This conserves power and extends battery life.
All four EL lamp-driving functions are on-chip. These are the switch-peak-to-peak. This conserves power and extends battery life.

## Key Features

- 112V peak-to-peak typical AC output voltage
- Low input current (w/ inductor current)...... 12 mA
- Low disabled input current......50nA
- Wide operating voltage range - from 2.0 V to 6.5 V
- Simple design requires few passive components
- Adjustable output lamp frequency controls lamp color and power consumption
- Adjustable converter frequency for minimum power consumption
- IMP803 pin-compatible
- MicroSO package option


## Applications

- N ight lights
- Automotive displays
- Cellular phones
- Pagers
- Clocks and radios
- Portable G PS receivers
- LCD module backlights


## Block Diagram



Pin Configuration


Pin Compatible With IMP803

## Ordering Information

| Part Number | Input Voltage | Regulated Output Voltage | Temperature Range | Pins-Package |
| :--- | :---: | :---: | :---: | :---: |
| IMP560EMA | 2.0 V to 6.5 V | YES | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $8-\mathrm{MicroSO}$ |
| IMP560ESA | 2.0 V to 6.5 V | YES | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | $8-\mathrm{SO}$ |
| IMP560/D | 2.0 V to 6.5 V | YES | $25^{\circ} \mathrm{C}$ | Dice |

Add /T to ordering part number for Tape and Reel.

## Absolute Maximum Ratings

| Supply Voltage, $\mathrm{V}_{\text {DD }}, \mathrm{V}_{\text {RSw-OSC }}$ and $\mathrm{V}_{\text {REL-OSC }}$ | -0.5 V to +7.0 V |
| :---: | :---: |
| Output Voltage, V ${ }_{\text {CS }}$ | -0.5 V to +120 V |
| Operating Temperature Range | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |
| Storage Temperature Range | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |
| Power Dissipation (SO) | 400 mW |
| Power Dissipation (MicroSO) | 300 mW |

Note: All voltages are referenced to GND.
These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability.

## Electrical Characteristics

Unless otherwise noted, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{R}_{\mathrm{SW}}=750 \mathrm{k} \Omega, \mathrm{R}_{\mathrm{EL}}=2.0 \mathrm{M} \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ON-resistance of MOS Switch | $\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}$ | $\mathrm{I}=100 \mathrm{~mA}$ |  | 3.5 | 8 | $\Omega$ |
| Output Voltage Regulation | $\mathrm{V}_{\text {CS }}$ | $\mathrm{V}_{\mathrm{DD}}=2.0$ to 6.5 V | 52 | 56 | 65 | V |
| Output Voltage Peak-to-peak (in regulation) | $\mathrm{V}_{\mathrm{A}}-\mathrm{V}_{\mathrm{B}}$ | $\mathrm{V}_{\mathrm{DD}}=2.0$ to 6.5 V | 104 | 112 | 120 | V |
| Quiescent $\mathrm{V}_{\text {DD }}$ Supply Current, Disabled | $\mathrm{I}_{\text {DDIS }}$ | $\mathrm{V}_{\text {RSW-OSC }}<100 \mathrm{mV}$ |  | 50 | 200 | nA |
| Input Current at $\mathrm{V}_{\text {DD }}$ Pin | $\mathrm{I}_{\mathrm{DD}}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 |  | 470 | 700 | $\mu \mathrm{A}$ |
| Input Current at $\mathrm{V}_{\mathrm{DD}}$ Pin | $\mathrm{I}_{\mathrm{DD}}$ | $V_{D D}=5.0 \mathrm{~V}$, See Figure 2 |  | 500 | 750 | $\mu \mathrm{A}$ |
| Input Current: $\mathrm{I}_{\mathrm{DD}}$ Plus Inductor Current | 1 N | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 |  | 12 |  | mA |
| $\mathrm{V}_{\text {A-B }}$ Output Drive Frequency | $\mathrm{f}_{\mathrm{EL}}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 | 300 | 370 | 430 | Hz |
| Switching Frequency | $\mathrm{f}_{\text {Sw }}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 | 50 | 70 | 90 | kHz |
| Switching Duty Cycle | $\mathrm{D}_{\text {SW }}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 |  | 88 |  | \% |

## Typical Characteristics






$l_{D D}$ vs. $V_{D D}$


## Pin Descriptions

| Pin Number | Name | Function |
| :---: | :---: | :---: |
| 1 | $\mathrm{V}_{\mathrm{DD}}$ | Positive voltage supply for the IMP560. Inductor L may be connected here or to a separate unregulated supply. |
| 2 | $\mathrm{R}_{\text {Sw-osc }}$ | Switch-mode resistor pin. Switching frequency is determined by an external resistor, $\mathrm{R}_{\text {Sw }}$. |
| 3 | $\mathrm{C}_{\text {S }}$ | Boost converter storage capacitor. The voltage across the EL lamp is equal to twice the voltage at $\mathrm{C}_{\mathrm{s}}$. |
| 4 | $L_{x}$ | Connection to flyback inductance, L. |
| 5 | GND | Ground pin. |
| 6 | $V_{B}$ | EL lamp drive. The lamp is connected in a high-voltage bridge circuit with $\mathrm{V}_{\mathrm{B}}$ providing the complementary connection to $\mathrm{V}_{\mathrm{A}}$. The peak-to-peak AC voltage across the EL lamp is thus two times $\mathrm{V}_{\mathrm{cs}}$. |
| 7 | $V_{\text {A }}$ | EL lamp drive. (See above) |
| 8 | $\mathrm{R}_{\text {EL-OSC }}$ | The EL lamp oscillator frequency setting pin. The oscillator frequency is controlled by external resistor $\mathrm{R}_{\mathrm{EL}}$. |

## External Components

| External Component | Description and Selection Guide |
| :---: | :---: |
| Diode | A fast reverse recovery diode, with BV > 100, such as a 1 N4148. |
| Capacitor $\mathrm{C}_{\text {s }}$ | This is the high voltage capacitor that stores the inductive energy transferred through the diode. A 100 volt capacitor between 10 nF and 100 nF is recommended. |
| Resistor $\mathrm{R}_{\mathrm{EL}}$ | The EL lamp oscillator frequency setting resistor. This resistor, connected between the $R_{E L-O s C}$ pin and ground, provides an oscillator frequency inversely proportional to $R_{E L}$; as $R_{E L}$ increases, the EL lamp frequency decreases along with the current drawn by the lamp. Lamp color is also determined by this frequency. A $2 \mathrm{M} \Omega$ resistor between the $\mathrm{R}_{\mathrm{EL} \text {-osc }}$ pin and the $V_{D D}$ supply results in a lamp frequency around 350 Hz : a $1 \mathrm{M} \Omega$ resistor will give $\approx 700 \mathrm{~Hz}$. |
| Resistor $\mathrm{R}_{\text {sw }}$ | Switching Oscillator frequency setting resistor. The switching oscillator resistor is connected between the $\mathrm{R}_{\mathrm{Sw} \text {-osc }}$ pin and the $\mathrm{V}_{\mathrm{DD}}$ supply. The switching frequency is inversely proportional to the resistor value, dropping as the resistance increases. |
| Inductor L | The inductor provides the voltage boost needed by means of inductive "flyback". The internal MOSFET switch alternately opens and closes the ground connection for the inductor at the $L_{x}$ pin. When this internal switch opens, the inductor potential will forward-bias the diode and the current will pass through the storage capacitor $\mathrm{C}_{\mathrm{s}}$, charging it to a high voltage. <br> Smaller inductors are preferred to prevent saturation. As the value of the inductor increases (and the series DC resistance of the inductor decreases), the switching frequency set by $R_{S W}$ should be increased to prevent saturation. In general, smaller value inductors that can handle more current are more desirable when larger area EL lamps must be driven. |

## High-Voltages Present

The IMP560 generates high voltages and caution should be exercised.

## Application Information

## Test and Application Circuit, 3.0V

Figure 1 shows the IMP560 configured to drive a 3-square-inch EL lamp, represented as a 10 nF capacitor.

Test and Application Circuit, 5.0V
Figure 2 shows a 5.0 V input application driving a 6 -square-inch EL lamp.


Figure 1. 3.0V Application


Figure 2. 5.0V Application

## Enable/ Disable Operation

Figure 3 shows the IMP560 can be enabled via a logic gate that connects $R_{S W}$ to $V_{D D}$, and disabled by connecting it to ground. $R_{E L}$ may be connected either to $\mathrm{V}_{\mathrm{DD}}$ or to the gate.

| Enable/Disable Table |  |
| :---: | :---: |
| $\mathbf{R}_{\text {SW }}$ Connection | IMP560 State |
| $V_{D D}$ | Enabled |
| Ground | Disabled |



Figure 3. Enable/Disable Operation

## Dual Supply Operation with 1.5V Battery

The IMP560 can also be operate from a single battery cell when a regulated voltage higher than 2.0 V is also available. The dual supply configuration, shown in Figure 4, uses the regulated voltage to operate the IMP560 while the energy for the highvoltage boost circuit comes from the battery. The current to run the internal logic is typically $420 \mu \mathrm{~A}$.

The circuit of Figure 4 can also be used with batteries that exceed 6.0 V as long as $\mathrm{V}_{\mathrm{DD}}$ does not exceed 6.5 V .


Figure 4. Dual Supply Operation with High Battery Voltages

## High-Voltage EL Lamp Driver

The IMP803 is an Electroluminescent (EL) lamp driver with the four EL lamp driving functions on-chip. These are the switch-mode power supply, its high-frequency oscillator, the high-voltage H-bridge lamp driver and its low-frequency oscillator. The IMP803 drives EL lamps of up to 30nF capacitance to high brightness; EL lamps with capacitances greater than 30 nF can be driven, but will be lower in light output. The typical regulated output voltage that is applied to the EL lamp is 180 V peak-topeak. The circuit requires few external components, a single inductor, single diode, two capacitors and three resistors. Two of these resistors set the frequency for two internal oscillators.

The IMP803 operates over a 2.0 V to 6.5 V supply voltage range. A regulated, low-power source can supply the low quiescent current of the IMP803. The inductor may be driven from an independent, unregulated supply voltage in dual supply applications.

An internal circuit shuts down the switching regulator when the lamp drive voltage reaches 180 V peak-to-peak. This conserves power and extends battery life.
The IMP803 is available in MicroSO and SO-8 packages and in die or wafer form.

Block Diagram

## Key Features

- Low Power: $420 \mu \mathrm{~A}$ typical $\mathrm{V}_{\mathrm{DD}}$ current
- Wide operating voltage range - from 2.0 V to 6.5 V
- 180V peak-to-peak typical AC output voltage
- Large output load capability - drive lamps with more than 30 nF capacitance
- Adjustable output lamp frequency for control of lamp color, lamp life, and power consumption
- Adjustable converter frequency to minimize power consumption
- Device can be Enabled/ Disabled
- Low quiescent current - 20nA (disabled)
- High-Voltage CMOS Process
- MicroSO package option


## Applications

- G PS units/ Pagers/ Cellular phones
- PDAs/ Handheld computers
- Safety illumination
- Portable instrumentation
- Battery-operated displays
- LCD modules
- Toys



## SO/ MicroSO



## Ordering Information

| Part Number | Input Voltage | Regulated Output Voltage | Temperature Range | Pins-Package |
| :--- | :---: | :---: | :---: | :---: |
| IMP803LG | 2.0 V to 6.5 V | Yes | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 -SO |
| IMP803IMA | 2.0 V to 6.5 V | Yes | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 8 -MicroSO |
| IMP803SX | 2.0 V to 6.5 V | Yes | $25^{\circ} \mathrm{C}$ | Dice |

Add /T to ordering part number for Tape and Reel.

## Absolute Maximum Ratings

$\mathrm{V}_{\text {DD }}, \mathrm{V}_{\text {RSW-OSC }}$ and $\mathrm{V}_{\text {ReL-OSC }}$. . . . . . . . . . . . . . .

Operating Temperature Range $\ldots \ldots . . . . . .-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Storage Temperature Range $\ldots \ldots \ldots \ldots, \ldots-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Power Dissipation (SO) $\qquad$
Power Dissipation (MicroSO) $\qquad$
$\mathrm{V}_{\mathrm{A}}, \mathrm{V}_{\mathrm{B}}$ 300 mW
-0.5 V to $\mathrm{V}_{\mathrm{CS}}(\mathrm{pin} 3)$

Note: All voltages are referenced to GND.
These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability.

## Electrical Characteristics

Unless otherwise noted, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{R}_{\mathrm{SW}}=750 \mathrm{k} \Omega, \mathrm{R}_{\mathrm{EL}}=2.0 \mathrm{M} \Omega$, and $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$.

| Parameter | Symbol | Conditions | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ON-resistance of MOS Switch | $\mathrm{R}_{\mathrm{DS} \text { (ON) }}$ | $\mathrm{I}=100 \mathrm{~mA}$ |  | 3.5 | 8 | $\Omega$ |
| Output Voltage Regulation | $\mathrm{V}_{\text {cs }}$ | $\mathrm{V}_{\mathrm{DD}}=2.0$ to 6.5 V | 80 | 90 | 100 | V |
| Output Voltage Peak-to-peak (in regulation) | $V_{A}-V_{B}$ | $\mathrm{V}_{\mathrm{DD}}=2.0$ to 6.5 V | 160 | 180 | 200 | V |
| Quiescent $\mathrm{V}_{\text {DD }}$ Supply Current, Disabled | IDDQ | $\mathrm{V}_{\text {RSW -osc }}<100 \mathrm{mV}$ |  | 20 | 200 | nA |
| Input Current at $V_{D D}$ Pin | $\mathrm{I}_{\mathrm{D}}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 |  | 420 | 700 | $\mu \mathrm{A}$ |
| Input Current at $\mathrm{V}_{\mathrm{DD}}$ Pin | $\mathrm{I}_{\mathrm{DD}}$ | $V_{D D}=5.0 \mathrm{~V}$, See Figure 2 |  | 500 | 750 | $\mu \mathrm{A}$ |
| Input Current: $\mathrm{I}_{\text {D }}$ Plus Inductor Current | $\mathrm{I}_{\mathrm{N}}$ | $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}$, See Figure 1 |  | 20 | 31 | mA |
| Output Voltage at $\mathrm{V}_{C S}$ | $\mathrm{V}_{\mathrm{cs}}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 | 60 | 74 | 100 | V |
| $\mathrm{V}_{A-B}$ Output Drive Frequency | $\mathrm{f}_{\text {EL }}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 | 300 | 370 | 430 | Hz |
| Switching Frequency | $\mathrm{f}_{\text {sw }}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 | 50 | 70 | 90 | kHz |
| Switching Duty Cycle | $\mathrm{D}_{\text {sw }}$ | $V_{D D}=3.0 \mathrm{~V}$, See Figure 1 |  | 88 |  | \% |


$\mathrm{V}_{\mathrm{CS}}, \mathrm{l}_{\mathrm{IN}}$ vs. $\mathrm{V}_{\text {IN }}$

$\mathrm{V}_{\mathrm{CS}}, \mathrm{I}_{\mathrm{IN}}$ vs. $\mathrm{V}_{\text {IN }}$

$I_{D D}$ vs. $V_{D D}$




## Pin Descriptions

| Pin Number | Name | Function |
| :---: | :---: | :---: |
| 1 | $V_{D D}$ | Positive voltage supply for the IMP803. Inductor L may be connected here or to a separate unregulated supply. |
| 2 | $\mathrm{R}_{\text {SW-OSC }}$ | Switch-mode resistor pin. Switching frequency is determined by an external resistor, $\mathrm{R}_{\text {Sw }}$. |
| 3 | $\mathrm{C}_{\mathrm{s}}$ | Boost converter storage capacitor. The voltage across the EL lamp is equal to twice the voltage at $\mathrm{C}_{\mathrm{s}}$. |
| 4 | $L_{x}$ | Connection to flyback inductance, L. |
| 5 | GND | Ground pin. |
| 6 | $V_{B}$ | EL lamp drive. The lamp is connected in a high-voltage bridge circuit with $\mathrm{V}_{\mathrm{B}}$ providing the complementary connection to $\mathrm{V}_{\mathrm{A}}$. The peak-to-peak AC voltage across the EL lamp is thus two times $\mathrm{V}_{\text {Cs }}$. |
| 7 | $V_{\text {A }}$ | EL lamp drive. (See above) |
| 8 | $\mathrm{R}_{\text {EL-OSC }}$ | The EL lamp oscillator frequency setting pin. The oscillator frequency is controlled by external resistor $\mathrm{R}_{\mathrm{EL}}$. |

## External Components

| External Component | Description and Selection Guide |
| :---: | :---: |
| Diode | Catch diode. A fast reverse recovery diode, with BV > 100, such as a 1 N4148. |
| Capacitor $\mathrm{C}_{\text {s }}$ | This is the high voltage capacitor that stores the inductive energy transferred through the catch diode. A 100 volt capacitor between 10 nF and 100 nF is recommended. |
| Resistor R ELL | The EL lamp oscillator frequency setting resistor. This resistor, connected between the $R_{E L-o s c}$ pin and $V_{D D}$, provides an oscillator frequency inversely proportional to $R_{E L}$; as $R_{E L}$ increases, the EL lamp frequency decreases along with the current drawn by the lamp. Lamp color is also determined by this frequency. A $2 \mathrm{M} \Omega$ resistor between the $\mathrm{R}_{\mathrm{EL} \text {-osc }}$ pin and the $V_{D D}$ supply results in a lamp frequency around 350 Hz : a $1 \mathrm{M} \Omega$ resistor will give $\approx 700 \mathrm{~Hz}$. |
| Resistor R ${ }_{\text {Sw }}$ | Switching Oscillator frequency setting resistor. The switching oscillator resistor is connected between the $R_{S W-O S C}$ pin and the $V_{D D}$ supply. The switching frequency is inversely proportional to the resistor value, dropping as the resistance increases. |
| Inductor L | The inductor provides the voltage boost needed by means of inductive "flyback". The internal MOSFET switch alternately opens and closes the ground connection for the inductor at the $L_{x}$ pin. When this internal switch opens, the inductor potential will forward-bias the catch diode and the current will pass through the storage capacitor $\mathrm{C}_{\mathrm{S}}$, charging it to a high voltage. <br> Smaller inductors are preferred to prevent saturation. As the value of the inductor increases (and the series DC resistance of the inductor decreases), the switching frequency set by $\mathrm{R}_{\mathrm{SW}}$ should be increased to prevent saturation. In general, smaller value inductors that can handle more current are more desirable when larger area EL lamps must be driven. |
| Lamp, $\mathrm{R}_{\mathrm{CL}}$ | An external resistor $\left(\mathrm{R}_{\mathrm{CL}}\right)$ in series with the lamp will protect the output drivers from high transient currents during lamp commutation. |

## Application Information

## Test and Application Circuit, 3.0V

Figure 1 shows the IMP803 configured to drive a 3-square-inch EL lamp, represented as a 10 nF capacitor. With a 3.0 V input, the EL lamp will be driven to moderate brightness.


Figure 1. 3.0V Application

## Test and Application Circuit, 5.0V

Figure 2 shows a 5.0 V input application driving a 6 -square-inch EL lamp.


Figure 2. 5.0V Application

## Test and Application Circut, 6.0V

At higher input voltage levels, the IMP803 will drive large EL lamps. Figure 3 shows a 6.0 V circuit configuration that will drive a 10 square-inch lamp.


1. Murata part \# LQH4N561K04 (DC resistance <14.5 $\Omega$ )
2. Larger values may be required depending upon supply impedance.

Figure 3. 6.0V Application

## Enable/ Disable O peration

Figure 4 shows that the IMP803 can be enabled via a logic gate that connects $\mathrm{R}_{\mathrm{SW}}$ to $\mathrm{V}_{\mathrm{DD}}$, and disabled by connecting it to ground.

| Enable/Disable Table |  |
| :---: | :---: |
| R $_{\text {SW }}$ Connection | IMP803 State |
| $\mathrm{V}_{\mathrm{DD}}$ | Enabled |
| GND | Disabled |



Note:

1. Murata part \# LQH4N561K04 (DC resistance <14.5 ת)
2. Larger values may be required depending upon supply impedance.

Figure 4. Enable/Disable Operation

## Dual Supply Operation with 1.5V Battery

The IMP803 can also be operate from a single battery cell when a regulated voltage higher than 2.0 V is also available. This dual supply configuration, shown in Figure 5, uses the regulated voltage to operate the IMP803 while the energy for the highvoltage boost circuit comes from the battery.

The circuit of Figure 5 thus allows operation with batteries that are below the 2 V minimum specification or above the 6.0 V maximum operating voltage.


Figure 5. Dual Supply Operation

## Switch Resistance

The IMP803 inductor switch resistance is typically below $3.5 \Omega$, as shown in Figure 6.


Figure 6. Boost Switch On Resistance

## High- Voltages Present

The IMP803 generates high voltages and caution should be exercised.

## IMP525

## Single Cell Battery Powered EL Lamp Driver

## General Information

| Die Thickness: | 25 mils (625 microns) |
| :--- | :--- |
| Bond Wire Size: | $1.0 \mathrm{mil}(25 \mathrm{microns})$ |
| Back Side Metal: | None |
| Back Side Potential: | Ground |
| Die Attach Method: | Conductive Adhesive |
| Bond Pad Metal: | Aluminum, $1 \%$ Silicon, $1 / 2 \%$ Copper |
| Bond Pad Size: | 100 microns per side |
| Die Size: | $1.35 \mathrm{~mm} \times 1.54 \mathrm{~mm}$ |



## Pad Description

| Pad <br> Number | Name | Function |
| :---: | :---: | :--- |
| 1 | $\mathrm{~V}_{\mathrm{DD}}$ | Positive voltage supply. |
| 2 | R $_{\text {SW-OsC }}$ | Switch-mode oscillator frequency setting pad. |
| 3 | $\mathrm{C}_{\mathrm{S}}$ | Boost converter storage capacitor pad. |
| 4 | $\mathrm{~L}_{\mathrm{X}}$ | Inductor pad. |
| 5 | GND | Ground pad. |
| 6 | $\mathrm{~V}_{\mathrm{B}}$ | EL lamp drive. |
| 7 | $\mathrm{~V}_{\mathrm{A}}$ | EL lamp drive. |
| 8 | REL-osc | EL lamp oscillator frequency setting pad. |
| $9^{*}$ | DIS | Disable pad. DIS = HIGH disables chip. |

* See Ordering Information table


## Pad Location ${ }^{1}$

| Pad <br> Number | X (microns) | Y (microns) |
| :---: | :---: | :---: |
| 1 | 1153 | 1092 |
| 2 | 476 | 1226 |
| 3 | 314 | 1226 |
| 4 | 143 | 1216 |
| 5 | 111 | 460 |
| 6 | 397 | 112 |
| 7 | 1104 | 112 |
| 8 | 1153 | 958 |
| 9 | 1153 | 1226 |

Notes 1. To bonding pad center

## Ordering Information

|  |  | Description |  |
| :---: | :---: | :---: | :---: |
| Part <br> Number | Pad <br> Number | Disable Pad <br> Active | Disable Pad <br> Not Active |
| IMP525/D | 9 |  | $\bullet$ |
| IMP525/D1 | 9 | $\bullet$ |  |

## IMP527

Single Cell Battery Powered EL Lamp Driver, 180V ${ }_{\text {PP }}$ Drive

## General Information

| Die Thickness: | 25 mils (625 microns) |
| :--- | :--- |
| Bond Wire Size: | 1.0 mil (25 microns) |
| Back Side Metal: | None |
| Back Side Potential: | Ground |
| Die Attach Method: | Conductive Adhesive |
| Bond Pad Metal: | Aluminum, $1 \%$ Silicon, $1 / 2 \%$ Copper |
| Bond Pad Size: | 100 microns per side |
| Die Size: | $1.35 \mathrm{~mm} \times 1.54 \mathrm{~mm}$ |



## Pad Description

| Pad <br> Number | Name | Function |
| :---: | :---: | :--- |
| 1 | $\mathrm{~V}_{\mathrm{DD}}$ | Positive voltage supply. |
| 2 | R $_{\text {Sw-osc }}$ | Switch-mode oscillator frequency setting pad. |
| 3 | $\mathrm{C}_{\mathrm{S}}$ | Boost converter storage capacitor pad. |
| 4 | $\mathrm{~L}_{\mathrm{X}}$ | Inductor pad. |
| 5 | GND | Ground pad. |
| 6 | $\mathrm{~V}_{\mathrm{B}}$ | EL lamp drive. |
| 7 | $\mathrm{~V}_{\mathrm{A}}$ | EL lamp drive. |
| 8 | R EL-osC | EL lamp oscillator frequency setting pad. |
| $9^{*}$ | DIS | Disable pad. DIS = HIGH disables chip. |

* See Ordering Information table


## Pad Location ${ }^{1}$

| Pad <br> Number | X (microns) | Y (microns) |
| :---: | :---: | :---: |
| 1 | 1153 | 1092 |
| 2 | 476 | 1226 |
| 3 | 314 | 1226 |
| 4 | 143 | 1216 |
| 5 | 111 | 460 |
| 6 | 397 | 112 |
| 7 | 1104 | 112 |
| 8 | 1153 | 958 |
| 9 | 1153 | 1226 |

Notes 1. To bonding pad center

## Ordering Information

|  | Disable <br> Part <br> Number | Pad <br> Number | Disable Pad <br> Active |
| :---: | :---: | :---: | :---: |
| IMP527/D |  |  | Disable Pad <br> Not Active |
| IMP527/D1 | 9 | $\bullet$ | $\bullet$ |

## iMP <br> ISO 9001 Registered

## IMP528

High-Voltage EL Lamp Driver, 220V PP Drive

## General Information

| Die Thickness: | 25 mils (625 microns) |
| :--- | :--- |
| Bond Wire Size: | $1.0 \mathrm{mil}(25$ microns) |
| Back Side Metal: | None |
| Back Side Potential: | Ground |
| Die Attach Method: | Conductive Adhesive |
| Bond Pad Metal: | Aluminum, $1 \%$ Silicon, $1 / 2 \%$ Copper |
| Bond Pad Size: | 100 microns per side |
| Die Size: | $1.38 \mathrm{~mm} \times 1.82 \mathrm{~mm}$ |



Pad Description

| Pad <br> Number | Name | Function |
| :---: | :---: | :---: |
| 1 | $V_{D D}$ | Positive voltage supply. |
| 2 | $\mathrm{R}_{\text {SW-OsC }}$ | Switch-mode oscillator frequency setting pad. |
| 3 | $\mathrm{C}_{S}$ | Boost converter storage capacitor pad. |
| 4 | $L_{x}$ | Inductor pad. |
| 5 | GND | Ground pad. |
| 6 | $V_{B}$ | EL lamp drive. |
| 7 | $V_{\text {A }}$ | EL lamp drive. |
| 8 | $\mathrm{R}_{\text {EL-OSC }}$ | EL lamp oscillator frequency setting pad. |
| 9 | GND | Ground pad. |
| 10* | DIS | Disable pad. DIS = HIGH disables chip. |

* See Ordering Information table


## Pad Location ${ }^{1}$

| Pad <br> Number | X (microns) | Y (microns) |
| :---: | :---: | :---: |
| 1 | 152 | 1480 |
| 2 | 152 | 1253.5 |
| 3 | 152 | 387.75 |
| 4 | 152 | 122.5 |
| 5 | 1198.5 | 140 |
| 6 | 1215 | 395 |
| 7 | 1215 | 1208.5 |
| 8 | 1234 | 1508.5 |
| 9 | 998 | 122.5 |
| 10 | 382 | 1553.5 |

Notes 1. To bonding pad center

## Ordering Information

|  | Disable <br> Part <br> Number | Pad <br> Number | Disable Pad <br> Active |
| :---: | :---: | :---: | :---: |
| IMP528/D |  |  | Disable Pad <br> Not Active |
| IMP528/D1 | 10 | $\bullet$ | $\bullet$ |

## IMP560

## Power Efficient EL Lamp Driver

## General Information

| Die Thickness: | 25 mils (625 microns) |
| :--- | :--- |
| Bond Wire Size: | 1.0 mil (25 microns) |
| Back Side Metal: | None |
| Back Side Potential: | Ground |
| Die Attach Method: | Conductive Adhesive |
| Bond Pad Metal: | Aluminum, $1 \%$ Silicon, $1 / 2 \%$ Copper |
| Bond Pad Size: | 100 microns per side |
| Die Size: | $1.38 \mathrm{~mm} \times 1.82 \mathrm{~mm}$ |



## Pad Description

| Pad <br> Number | Name | Function |
| :---: | :---: | :--- |
| 1 | $\mathrm{~V}_{\mathrm{DD}}$ | Positive voltage supply. |
| 2 | R $_{\text {SW-OSC }}$ | Switch-mode oscillator frequency setting pad. |
| 3 | $\mathrm{C}_{\mathrm{S}}$ | Boost converter storage capacitor pad. |
| 4 | $\mathrm{~L}_{\mathrm{X}}$ | Inductor pad. |
| 5 | GND | Ground pad. |
| 6 | $\mathrm{~V}_{\mathrm{B}}$ | EL lamp drive. |
| 7 | $\mathrm{~V}_{\mathrm{A}}$ | EL lamp drive. |
| 8 | $\mathrm{R}_{\mathrm{EL} \text {-OSC }}$ | EL lamp oscillator frequency setting pad. |
| 9 | GND | Ground pad. |
| $10^{*}$ | DIS | Disable pad. DIS = HIGH disables chip. |

* See Ordering Information table


## Pad Location ${ }^{1}$

| Pad <br> Number | X (microns) | Y (microns) |
| :---: | :---: | :---: |
| 1 | 152 | 1480 |
| 2 | 152 | 1253.5 |
| 3 | 152 | 387.75 |
| 4 | 152 | 122.5 |
| 5 | 1198.5 | 140 |
| 6 | 1215 | 395 |
| 7 | 1215 | 1208.5 |
| 8 | 1234 | 1508.5 |
| 9 | 998 | 122.5 |
| 10 | 382 | 1553.5 |

Notes 1. To bonding pad center

## Ordering Information

|  |  | Description |  |
| :---: | :---: | :---: | :---: |
| Part <br> Number | Pad <br> Number | Disable Pad <br> Active | Disable Pad <br> Not Active |
| IMP560/D | 10 |  | $\bullet$ |
| IMP560/D1 | 10 | $\bullet$ |  |

## IMP803

## High-Voltage EL Lamp Driver

## General Information

| Die Thickness: | 25 mils (625 microns) |
| :--- | :--- |
| Bond Wire Size: | $1.0 \mathrm{mil}(25$ microns) |
| Back Side Metal: | None |
| Back Side Potential: | Ground |
| Die Attach Method: | Conductive Adhesive |
| Bond Pad Metal: | Aluminum, $1 \%$ Silicon, $1 / 2 \%$ Copper |
| Bond Pad Size: | 100 microns per side |
| Die Size: | $1.38 \mathrm{~mm} \times 1.82 \mathrm{~mm}$ |



## Pad Description

| Pad <br> Number | Name | Function |
| :---: | :---: | :---: |
| 1 | VDD | Positive voltage supply. |
| 2 | RSw-osc | Switch-mode oscillator frequency setting pad. |
| 3 | $\mathrm{C}_{s}$ | Boost converter storage capacitor pad. |
| 4 | Lx | Inductor pad. |
| 5 | GND | Ground pad. |
| 6 | $V_{B}$ | EL lamp drive. |
| 7 | $V_{\text {A }}$ | EL lamp drive. |
| 8 | $\mathrm{R}_{\text {EL-OSC }}$ | EL lamp oscillator frequency setting pad. |
| 9 | GND | Ground pad. |
| 10* | DIS | Disable pad. DIS = HIGH disables chip. |

* See Ordering Information table


## Pad Location ${ }^{1}$

| Pad <br> Number | X (microns) | Y (microns) |
| :---: | :---: | :---: |
| 1 | 152 | 1480 |
| 2 | 152 | 1253.5 |
| 3 | 152 | 387.75 |
| 4 | 152 | 122.5 |
| 5 | 1198.5 | 140 |
| 6 | 1215 | 395 |
| 7 | 1215 | 1208.5 |
| 8 | 1234 | 1508.5 |
| 9 | 998 | 122.5 |
| 10 | 382 | 1553.5 |

Notes 1. To bonding pad center

## Ordering Information

|  |  | Description |  |
| :---: | :---: | :---: | :---: |
| Part <br> Number | Pad <br> Number | Disable Pad <br> Active | Disable Pad <br> Not Active |
| IMP803SX | 10 |  | $\bullet$ |
| IMP803/D1 | 10 | $\bullet$ |  |

## Notes



# Electroluminescent Lamp Driver Evaluation Board 

## Introduction

This Application Note introduces an Evaluation Board for IMP EL driver ICs. It is supplied with the IMP803 but can also be used with the IMP560 and IMP525: all 3 have identical pinouts.

## EL Lamps and Drivers

An electroluminescent (EL) lamp consists of a phosphor coating on a dielectric that is sandwiched between two conductors. Electrically, it looks like a capacitor. Such a lamp requires drive from a high alternating voltage source in order to emit light. This can be obtained from IMP integrated circuits IMP803, IMP560 and IMP525 that convert low voltages into appropriate high-voltage waveforms.

Small EL lamps exhibit about 2 to $6 \mathrm{nF} / \mathrm{in}^{2}$. IMP Driver ICs are capable of powering EL lamps that have total equivalent load capacitances up to 30 nF , so this works out to a maximum of around 15 square inches. "Powering" in this context means enabling enough light for the application, which can range from LCD backlights (relatively bright in a handheld device) to pagers (medium-bright, in a poorly-lit room), to night-lights (faint, in a dark room).

## IMP Driver IC System Diagram

As shown in Figure 1, these ICs contain a high-voltage MOSFET switch, an output H-bridge, and oscillators to drive each. The switch, combined with an external inductor and diode, form a step-up (boost) converter that transforms the input voltage to 4590 volts across capacitor CS. This, in turn, is switched from one side of the load (the EL lamp) to the other by a commutating bridge, driven by its own oscillator. This action causes the lamp to experience twice the $\mathrm{C}_{S}$ value (i.e. 90-180 volts peak-to-peak) with no DC component.

A typical application uses a switch frequency of 80 kHz and bridge commutation frequency of 360 Hz . These frequencies are controllable via external resistors; $R_{S W}$ for the boost converter and $R_{E L}$ for the output driver. $R_{E L}$ influences brightness, color and EL lamp life. $\mathrm{R}_{\mathrm{SW}}$ controls converter efficiency. Both affect power consumption.

IMP Driver IC System Diagram


Figure 1. Circuitry in gray is on-chip.

## Driver Variations

The IMP803, 560 and 525 have an internal regulating circuit (see Figure 2), that is useful where $\mathrm{V}_{\mathrm{IN}}$ is expected to change considerably, as with an aging battery: as $\mathrm{V}_{\text {IN }}$ falls, $\mathrm{V}_{\text {OUT }}$ (and brightness) will remain substantially unaffected.

Table 1 is a general comparison of IMP EL Lamp drivers. It facilitates choices based on number of batteries, size of display, and regulation. Required display brightness will also need to be factored into the choice.

Table 1. General Characteristics of IMP EL Lamp Drivers

| Device | $\mathbf{V}_{\text {IN }}$ | V OUT | Regulated Output | Max. Switch R(on) |
| :---: | :---: | :---: | :---: | :---: |
| IMP803 | $2.4-6.5 \mathrm{~V}$ | $180 \mathrm{~V}_{\text {PP }}$ | Yes | $8 \Omega$ |
| IMP560 | $2.4-6.5 \mathrm{~V}$ | $120 \mathrm{~V}_{\text {PP }}$ | Yes | $8 \Omega$ |
| IMP525 | $0.9-2.5 \mathrm{~V}$ | $112 \mathrm{~V}_{\text {PP }}$ | Yes | $15 \Omega$ |



Figure 2. Block Diagram for IMP circuits. Dotted components are equivalent to regulation circuitry (see text).

## Basic Circuit, Plus Variations

In normal operation, $\mathrm{V}_{\mathrm{DD}}$ is one or two 1.5 V cells and L 1 is a tiny ferrite-bobbin inductor. $\mathrm{R}_{\mathrm{SW}}$ and $\mathrm{R}_{\mathrm{EL}}$ control their respective oscillators. If a logic-controllable shutdown is desired, $R_{\text {SW }}$ may be switched between $V_{D D}$ and GND ( $\mathrm{I}_{\mathrm{DDQ}}=1 \mu \mathrm{~A}$ max.). Conversely, if shutdown is via $V_{D D}, R_{S W}$ should then be connected to $V_{D D}$ as shown by the dotted line in Figure 3.
$\mathrm{R}_{\mathrm{CL}}$ is included to protect the bridge against peak currents during commutation. A value of $500 \Omega$ to $2 \mathrm{k} \Omega$ is suitable.

In use, the inductor current can reach several tens of milliamperes, so in single-battery applications it is recommended that the low-current shutdown capability of the driver IC be utilized. This is done by connecting $\mathrm{R}_{\text {SW }}$ (point A on the schematics) to either $\mathrm{V}_{\mathrm{DD}}(\mathrm{ON})$ or GND (OFF). With power source(s) connected, shutdown (standby) current is typically much lower than $1 \mu \mathrm{~A}$.


Figure 3. Basic EL Lamp Driver.

## Reducing Component Count

Having said that keeping $R_{C L}$ is a good idea, it is true that removing as many components as possible may also be desirable. For the IMP803, $\mathrm{R}_{\mathrm{EL}}$ and $\mathrm{R}_{\mathrm{SW}}$ may be combined as shown in Figure 4. Varying $R_{\text {EL }}$ causes a visible change in brightness and color, but a similar variation in $R_{S W}$ (affecting oscillator frequency and power consumption) is much less noticeable. Combining the two is thus
a valid way to save a resistor. The bypass capacitor C $_{\text {BP }}$ (IMP525 only) reduces display flicker in noisy environments, such as when there is no ground plane.


Figure 4. Using $R_{S W}$ to supply current for both switch and EL oscillators, and also serve as a low-current on/off switch (IMP803 only).

Using the circuit in Figure 5, one can utilize an available $\mathrm{V}_{\text {IN }}$ that is higher or lower than the allowable $\mathrm{V}_{\mathrm{DD}}$. The logic shutdown may also be separated from $\mathrm{V}_{\mathrm{DD}}$. Such arrangements are helpful when the inductor supply is too low for the IC, or the display size requires a voltage that is too high for the IC.

A higher $\mathrm{V}_{\text {IN }}$ will need a higher switching frequency to keep the inductor out of saturation. In all cases, note the presence of HIGH VOLTAGE!


Figure 5. General Circuit, where chip $V_{D D}$, on/off logic and $V_{I N}$ are all different.

## Evaluation Board

The ELD002 is a PC board for evaluation and experimentation purposes. More compact arrangements are easily achieved by using surface-mounted components exclusively. The various possible connections mirror the options discussed in the data sheet and the Application Note. While the IMP803 is supplied on the board, other pin-compatible drivers may be substituted.

The two dark patches are the connections for the EL lamp which are made using conductive double-sided tape. The display itself is held down with ordinary double-sided tape. Taping is advantageous for several reasons, among which are that lamps with
staked connecting terminals generally cost more, and they are a possible site for mechanical (and thus electrical) failure.

As a general precaution, note that HIGH VOLTAGE exists on the board; around 180 V or so. The current level is low so there is no danger, except possible pain if a tender skin area or open cut contacts the HV sections.

There are extra holes for capacitors (if needed), and the hole spacings are wide enough to accommodate $1 / 4 \mathrm{~W}$ resistors. Corner mounting holes have also been provided.


Figure 6. Evaluation Board Layout and Schematic.

## Some Battery Considerations

To keep the board light in weight, a Li-Mn power source was selected. When energized, the drain from the circuit is around 22 mA , thus the CR battery chemistry is preferred over the BR for its superior pulse performance. If long-term continual illumination is anticipated and space is not an issue, alkaline batteries may be more economical.

With the IMP803 and $560 \mu \mathrm{H}$ inductor supplied, regulation begins at about $3-3.5 \mathrm{~V}$, but display illumination appears virtually
unchanged above 2.7 V . When choosing the battery chemistry, it is a good idea to match the cell "plateau" voltages to this. For example, a typical NiCad plateau is 1.2 V under load, so more than 2 cells would be needed. Alkaline plateaus are somewhat higher, and they differ with size, shape and duty, so 2 cells could suffice. Li-Mn coin cells have their voltage plateau under load at about 2.85 volts. They can drop lower, but they also return to close to 3 V when the load is removed.

## Additional Points

1) To experiment with the Figure 4 scheme, a jumper may be run from the rightmost pad of $R_{E L}$ to the leftmost pad of $C_{B P}$ (with the + above it). Start with an $R_{S W}$ of $750 \mathrm{k} \Omega$. Short leads and a ground plane are more critical in this arrangement.
2) $C_{S}$ should be $10 \mathrm{nF}-100 \mathrm{nF}$.
3) The IMP803, IMP560 and IMP525 datasheets show performance with different inductors. For example, high-voltage regulation is reached earlier with lower L, but this requires more current. This may be partially offset by adjustment of the oscillator resistors.
4) To experiment with multiple supplies, the appropriate jumpers may be removed.
5) The inclusion of $\mathrm{R}_{\mathrm{CL}}$ should be stressed: while $500 \Omega$ to $10 \mathrm{k} \Omega$ has been used, $2 \mathrm{k} \Omega$ is the best all-around value.

## Layout Rules for Other Arrangements

1) A ground plane is recommended to keep stray high frequencies confined. In a very small area, the need for a ground plane may be nil. A totally surface-mount arrangement would make such a plane difficult anyway.
2) Locate high voltages away from the high-impedance elements $R_{E L}$ and $R_{S W}$.
3) Make sure that $C_{S}$ has a rating of at least 100 V .
4) The diode should have good reverse-recovery characteristics (the general-purpose 1 N 4148 is adequate) and should be rated for pulsed BV $>100 \mathrm{~V}$ for the IMP803, and pulsed BV $>75 \mathrm{~V}$ for the IMP560 and IMP525.
5) Shutdown by a logic-level signal is possible by connecting $R_{S W}$ to ground ( $\mathrm{R}_{\mathrm{SW}}$ is normally connected to $\mathrm{V}_{\mathrm{DD}}$ ). This on/off logic uses only $1 \mu \mathrm{~A}$ max. when connected at this location.
6) Required voltage ratings for the capacitors other than $C_{S}$ are flexible, and need only reflect actual stresses plus a safety margin.

## Bill of Materials for ELD001

| Component | Description | Manufacturer | Part Number |
| :--- | :--- | :--- | :--- |
| Resistors ( $\pm 5 \%)$ | See Table, below |  |  |
| Capacitors ( $\pm 20 \%)$ | See Table, below | Murata | RPE121/122 Series |
| Switch | SPST, momentary | Panasonic | P8008S |
| Battery | 3.0 V Li-Mn Coin | Sony <br> Panasonic | CR2450-HE4 |
| Inductor | L1 $=560 \mu \mathrm{H}$ | Murata | LQH4N561K04 |
| Diode | D1 $=1 \mathrm{N4148}$ |  |  |
| Lamp | $1.3 " \times 2.05 "$ | MetroMark or other | ARclad 8001 |
| Conductive Tape | Connects display | Adhesives Research | Type 665 |
| Double-Sided Tape | Holds display down | $3 M$ |  |

## Key to Components and Ratings

| Component | Value | Function | Comments |
| :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\text {SW }}$ | $30 \mathrm{k} \Omega$ to $3 \mathrm{M} \Omega$ | Sets switch osc. frequency. | Decrease R to increase frequency. |
| $\mathrm{R}_{\mathrm{EL}}$ | $500 \mathrm{k} \Omega$ to $10 \mathrm{M} \Omega$ | Sets bridge osc. frequency. | Decrease R to increase frequency. |
| $\mathrm{R}_{\mathrm{CL}}$ | $500 \Omega$ to $2 \mathrm{k} \Omega$ | Limits output current. | Protects IC. |
| $\mathrm{C}_{\mathrm{s}}$ | $0.01 \mu \mathrm{~F}$ to $0.1 \mu \mathrm{~F}, 100 \mathrm{~V}$ | Stores high voltage. | Use low values for large lamps. |
| $\mathrm{C}_{\text {BATT }}$ | $0.1 \mu \mathrm{~F}, 10 \mathrm{~V}$ | Supply bypass. | Keeps supply impedance low. |
| $\mathrm{C}_{\mathrm{BP}}$ | 1nF, 10V | Lowers noise at $\mathrm{R}_{\text {Sw }}$. | IMP525 only. |
| $\mathrm{C}_{\text {IN }}$ | $0.1 \mu \mathrm{~F}$ to $22 \mu \mathrm{~F}$ | Supply bypass. | Keeps supply impedance low. |
| L1 | $100 \mu \mathrm{H}$ to 1 mH | Stores energy. | Small L, high f increases $\mathrm{V}_{\text {Out }}$. |
| D1 | 100V, 10mA (1N4148) | Passes energy from L to $\mathrm{C}_{\mathrm{s}}$. | Use fast recovery type. |

## APPEN DIX: Introduction to EL Lamps

Chemical compounds, called phosphors, glow when energy is applied to them. This excitation energy can come from conducted or radiated electrons, or an electric field. A common example of this process is found in the emitted (radiated) electrons that impinge on the dots and stripes of color monitors and TVs, whose phosphors emit everything from pure colors to white light, depending on their formulations.

Backlights and lamps generally are simpler, employing a manganese-activated zinc sulfide phosphor ( $\mathrm{ZnS}: \mathrm{Mn}$ ) that is excited by a high-voltage ( $>40 \mathrm{~V}$ ) AC electric field (DC can shorten the lamp life). Fabrication involves depositing the phosphor as a thin film onto a BaTiO3 dielectric between conducting planes, like a capacitor: one of the planes is the transparent conductor, indium tin oxide (ITO). The lamp color depends on phosphor formulation, but also on its physical realization (i.e. encapsulation, resins, dyes, etc.), plus the characteristics of the drive circuitry.

The IMP line of drivers is targeted mainly at applications like backlight EL and stand-alone pre-printed or segmented lamps. Backlights are used with the Liquid-Crystal Displays found in cellular telephones, pagers, Personal Digital Assistants (PDAs), and general-purpose local lighting applications where low power consumption without heat is important (e.g. airline cockpits, medical instrumentation).
The excitation required for lamps ranges from tens to hundreds of volts, at frequencies from 60 Hz to a few kHz . Each display has an optimum combination depending on size, color, efficiency and desired brightness.
In general, the changes in brightness with frequency and voltage are nearly linear. These facts allow tradeoffs. For example, if going above a certain voltage is not allowed, an increase in drive frequency may achieve the same result.

## Addendum

The new evaluation board ELD002 is now available and will be sent out to all new purchasers. This Addendum will serve to explain the differences and update the information in $\mathrm{AN}-1$.

## Changes

The diagrams below show the basic wiring of ELD001 (see AN-1, Figure 6) on the left and ELD002 on the right. The difference is that, with ELD002, the 3 V cell is switched to the $\mathrm{V}_{\mathrm{DD}}$ pin, and this voltage only goes to the inductor if JP1 is connected. If JP1 is open, an alternate voltage can be used to power the inductor.


ELD001


In contrast, the ELD001 switched power to the inductor, and to $\mathrm{V}_{\mathrm{DD}}$ only if JP1 was connected. This was intended to demonstrate the logic-level shutdown ability of the IMP803: by using the pushbutton for the heavy current to the inductor, the $\mathrm{V}_{\mathrm{DD}}$ pin could be tied to a voltage source and the chip enabled/disabled by a logic level of $\mathrm{V}_{\mathrm{DD}}$ or ground applied to $\mathrm{R}_{\mathrm{SW}}$. The ELD002 board allows both features to be exercised. For ELD001 users who wish to modify their boards, the changes are shown below.

## ELD002

## Updates

1) $I_{D D Q}$ is listed as $1 \mu A \max$. (AN-1, pp 3, 6, 7). Extensive testing has shown this to be much too conservative: $25 n A$ is much more typical.
2) In using $R_{S W}$ to shut-down the IMP803 (only $25 n A$ ), $R_{E L}$ can remain connected to the $V_{D D}$ pin; only 1 resistor then needs to be switched.
3) Under some circumstances, $R_{C L}$ can be omitted. Consult IMP for details.
4) For single-battery systems (the vast majority), the capacitor shown on the diagram as $C_{I N}$ is not needed. For cases where it is needed, further "surgery" is required: cut the trace shown as (A) and reconnect it as per the dotted line.

## EL Driver Demonstration PC Boards, IMP-DBM and IMP-DBS

## Introduction

These Demonstration Boards provide a platform for demonstration and experimentation with IMP's EL lamp drivers IMP803, IMP560 and IMP525. The PC board has space for all of the components required for a complete application circuit. In addition, compact size facilitates their use in prototype systems.

For normal operation, the enable pad (EN), the $\mathrm{V}_{\mathrm{DD}}$ pad and the $\mathrm{V}_{\mathrm{L}}$ pad are all connected to the positive supply voltage. If the board is located far from the supply, a $10 \mu \mathrm{~F} / 10 \mathrm{~V}$ tantalum capacitor from $\mathrm{V}_{\mathrm{L}}$ to GND should be used to keep supply impedance low (This cap, or its equivalent, is normally present in a manufactured circuit). Also, better noise immunity may be achieved by utilizing separate wires for the $\mathrm{V}_{\mathrm{L}}$ and $\mathrm{V}_{\mathrm{DD}}$ connections.
The $\mathrm{C}_{\text {BATT }}$ capacitor is used to bypass the supply pin of the IC. The $\mathrm{C}_{\text {SW }}$ capacitor (IMP525 only) is utilized to reduce noise on the high impedance $R_{S W}$ pin. $C_{S W}$ should never be greater than 100 pF since this can result in instability of the 525's internal oscillator.

The layout was designed to reduce the effects of noise through use of a ground plane and by separation of the high-current components (inductor, diode, and reservoir capacitor) from the high-impedance portion of the circuit (the high-value frequencysetting resistors). Additionally, the lengths of high-current traces were minimized.

If parts are replaced or exchanged by hand- soldering, care should be taken to thoroughly clean the residual flux from the board surface. Otherwise, resultant leakage currents may prevent proper operation of the part. The tight spacing and high impedances of input nodes on the PCB exacerbate this effect. The predominant impact of PCB leakage is a shift in the switch and commutation frequencies away from their designed values due to leakage currents from the $R_{S W}$ and $R_{E L}$ pins.



Figure 2. Demonstration Board Schematic.

Table 1. Bill of Materials (use as required)

| Component | Package | Manufacturer and Part Number | IMP803 | IMP560 | IMP525 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\mathrm{SW}}$ | 0603 | Any | $750 \mathrm{k} \Omega$ | $750 \mathrm{k} \Omega$ | $1 \mathrm{M} \Omega$ |
| $\mathrm{R}_{\mathrm{EL}}$ | 0603 | Any | $2 \mathrm{M} \Omega$ | $2 \mathrm{M} \Omega$ | $1 \mathrm{M} \Omega$ |
| $\mathrm{R}_{\mathrm{CL}}$ | 0603 | Any | $510 \Omega$ | - | - |
| L 1 | 1812 | Murata LQH4N561K04 | $560 \mu \mathrm{H}$ | $560 \mu \mathrm{H}$ | $560 \mu \mathrm{H}$ |
| $\mathrm{C}_{S}$ | 0805 | NovaCap 0805B683K101NT | $68 \mathrm{nF} / 100 \mathrm{~V}$ | $68 \mathrm{nF} / 100 \mathrm{~V}$ | $68 \mathrm{nF} / 100 \mathrm{~V}$ |
| $\mathrm{D} 1^{\mathrm{C}_{\mathrm{BATT}}}$ | SOD80 | $4148-$ type | 100 V | 75 V | 75 V |
| $\mathrm{C}_{S W}$ | 0603 | Any | Any | 100 nF | 100 nF |

Table 2. Component Description Table

| Component | Function | Comments |
| :---: | :---: | :---: |
| $\mathrm{R}_{\text {SW }}$ | Sets switch frequency | Decreasing R increases frequency. |
| $\mathrm{R}_{\mathrm{EL}}$ | Sets commutation frequency | Decreasing R increases frequency. |
| $\mathrm{R}_{\mathrm{CL}}$ | Limits output current | Optional external part: protects bridge if $\mathrm{V}\left(\mathrm{C}_{S}\right)>80 \mathrm{~V}$ (IMP803 only). |
| L1 | Boost inductor | Delivers energy to $\mathrm{C}_{\mathrm{s}}$. |
| $\mathrm{C}_{\text {S }}$ | Reservoir capacitor | Delivers energy to commutating bridge. |
| $\mathrm{C}_{\text {SW }}$ | Noise reduction capacitor | Optional, use if flickering is observed (IMP525 only). |
| $\mathrm{C}_{\text {BATt }}$ | Supply bypass capacitor | Optional (use if missing from external circuit) |
| D1 | Catch diode | Fast recovery diode recommended. Observe $\mathrm{BV}_{\text {REV }}$. |

## Package Dimensions

MicroSO (8-Pin)


MicroSO (8-Pin).eps

| Parts/Reel | 3000 |
| :--- | :--- |

SO (8-Pin)

Parts/Reel


| Inches |  |  | Millimeters |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Min | Max | Min | Max |
| MicroSO (8-Pin) |  |  |  |  |
| A | - | 0.0433 | - | 1.10 |
| A1 | 0.0020 | 0.0059 | 0.050 | 0.15 |
| A2 | 0.0295 | 0.0374 | 0.75 | 0.95 |
| b | 0.0098 | 0.0157 | 0.25 | 0.40 |
| C | 0.0051 | 0.0091 | 0.13 | 0.23 |
| D | 0.1142 | 0.1220 | 2.90 | 3.10 |
| e | 0.0256 BSC |  | 0.65 BSC |  |
| E | 0.193 BSC |  | 4.90 BSC |  |
| E1 | 0.1142 | 0.1220 | 2.90 | 3.10 |
| L | 0.0157 | 0.0276 | 0.40 | 0.70 |
| a | $0^{\circ}$ | $6^{\circ}$ | $0^{\circ}$ | $6^{\circ}$ |
| SO (8-Pin) |  |  |  |  |
| A | 0.053 | 0.069 | 1.35 | 1.75 |
| A1 | 0.004 | 0.010 | 0.10 | 0.25 |
| B | 0.013 | 0.020 | 0.33 | 0.51 |
| C | 0.007 | 0.010 | 0.19 | 0.25 |
| e | 0.050 |  | 1.27 |  |
| E | 0.150 | 0.157 | 3.80 | 4.00 |
| H | 0.228 | 0.244 | 5.80 | 6.20 |
| L | 0.016 | 0.050 | 0.40 | 1.27 |
| D | 0.189 | 0.197 | 4.80 | 5.00 |

## Tape Schematic ${ }^{5}$



For Tape Feeder Reference Only Including Draft and Radii Concentric Around BO


Embossed Tape - Constant Dimensions

| Tape Size | $\mathbf{D}$ | $\mathbf{E}$ | P0 | P2 | T Max. | T1 Max. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 mm | $1.5_{-0.0}^{+0.10}$ | $1.75 \pm 0.10$ | $4.0 \pm 0.10$ | $2.0 \pm 0.05$ | 0.600 | 0.10 |
| and | 0.0 <br> 12 mm | $\left.\begin{array}{c}0.004 \\ -0.0\end{array}\right)$ | $(0.069 \pm 0.004)$ | $(0.157 \pm 0.004)$ | $(0.079 \pm 0.002)$ | $(0.024)$ |

ELD/B_102.at3

## Embossed Tape - Variable Dimensions

| Tape Size | A0, B0, K0 | B1 <br> See Note 4 | $\begin{gathered} \text { D1 } \\ \text { See Note } 3 \end{gathered}$ | F | T2 | P1 | W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 mm 1/2 Pitch | See Note 1 | $\begin{gathered} 4.55 \\ (0.179) \end{gathered}$ | $\begin{gathered} 1.0 \\ (0.039) \end{gathered}$ | $\begin{gathered} 3.5 \pm 0.05 \\ (0.138 \pm 0.002) \end{gathered}$ | $\begin{aligned} & \text { 2.5 Max. } \\ & (0.098) \end{aligned}$ | $\begin{gathered} 2.0 \pm 0.10 \\ (0.079 \pm 0.004) \end{gathered}$ | $8.0 \begin{gathered}+0.3 \\ -0.1\end{gathered}$ |
| 8 mm |  |  |  |  |  | $\begin{gathered} 4.0 \pm 0.10 \\ (0.157 \pm 0.004) \end{gathered}$ | $\binom{0.315^{+0.012}}{-0.004}$ |
| 12mm | See Note 1 | $\begin{gathered} 8.2 \\ (0.323) \end{gathered}$ | $\begin{gathered} 1.5 \\ (0.059) \end{gathered}$ | $\begin{gathered} 5.5 \pm 0.05 \\ (0.217 \pm 0.002) \end{gathered}$ | $\begin{aligned} & \text { 6.5 Max. } \\ & (0.256) \end{aligned}$ | $\begin{gathered} 4.0 \pm 0.10 \\ (0.157 \pm 0.004) \\ \hline \end{gathered}$ | $\begin{gathered} 12.0 \pm 0.30 \\ (0.472 \pm 0.012) \end{gathered}$ |
| 12 mm <br> Double Pitch |  |  |  |  |  | $\begin{gathered} 8.0 \pm 0.10 \\ (0.315 \pm 0.004) \\ \hline \end{gathered}$ |  |

Notes: 1. $A 0, B 0$ and $K 0$ are determined by the maximum dimensions to the ends of the terminals extending from the component body and / or the body dimensions of the component. The clearance between the ends of the terminals or body of the component to the sides and depth of the cavity ( $\mathrm{A} 0, \mathrm{BO}$ and K 0 ) must be within $0.05 \mathrm{~mm}(0.002)$ minimum and $0.50 \mathrm{~mm}(0.020)$ maximum. The clearance allowed must also prevent rotation of the component within the cavity of not more than 20 degrees (see Component Rotation).
2. Tape with components shall pass around radius.
3. The embossment hole location shall be measured from the spocket hole controlling the location of the embossment. Dimensions of embossment location and hole location shall be applied independent of each other.
4. B1 dimension is a reference dimension for tape feeder clearance only.
5. Electronic Industries Association, Standard EIA-481-1.

## Tape Layout

## Component Rotation



Bending Radius


Emboss04.eps

Tape Camber (Top View)


Allowable camber to be $1 \mathrm{~mm} / 100 \mathrm{~mm}$ nonaccumulative over 250 mm .
Emboss05.eps

Tape Leader and Trailer Dimensions


Reel Dimension


| Tape Size | A Max. | B Min. | C | D Min. | N Min. | W1 | W2 Max. | W3 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 8 mm | $\begin{gathered} 330 \\ (12.992) \end{gathered}$ | $\begin{gathered} 1.5 \\ (0.059) \end{gathered}$ | $\begin{gathered} 13.0 \pm 0.20 \\ (0.512 \pm 0.008) \end{gathered}$ | $\begin{gathered} 20.2 \\ (0.795) \end{gathered}$ | $\begin{gathered} 50 \\ (1.969) \end{gathered}$ | $\left.\begin{array}{c} 8.4+1.5 \\ -0.0 \\ (0.331+0.059 \\ -0.0 \end{array}\right)$ | $\begin{gathered} 14.4 \\ (0.567) \end{gathered}$ | $\begin{gathered} \text { 7.9 Min. } \\ \text { (0.311) } \\ \text { 10.9 Max. } \\ (0.0429) \end{gathered}$ |
| 12 mm |  |  |  |  |  | $\begin{gathered} 12.4+2.0 \\ -0.0 \\ \binom{0.488+0.078}{-0.0} \end{gathered}$ | $\begin{gathered} 18.4 \\ (0.724) \end{gathered}$ | $\begin{gathered} \text { 11.9 Min. } \\ \text { (0.469) } \\ \text { 15.4 Max. } \\ (0.607) \end{gathered}$ |

## Tape Layout



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Web: www.kruvand.com
Austin, TX
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REPLIST 6-11-99

## From Oakland International Airport

Go South on 880 and turn right at the Montague Expressway exit, move left out of the car pool lane. Turn left on Zanker R oad and then turn right on Daggett Drive.

From San Francisco International Airport
Go South on 101 to the Montague Expressway exit (east). Turn right on Zanker R oad and then turn right


For Additional Directions
408-432-9100

From San Jose International Airport
From Terminal Drive go to Airport Blvd. From A irport Blvd., turn onto Airport Pkwy. (Airport Pkwy becomes Brokaw R oad after 101). Turn left on North First Street, then turn right on Daggett Drive.

## Quality Priority

Quality in everything we do is a fundamental IMP commitment. Quality may not be sacrificed for any other priority. Before any action is taken, the effect on quality as seen by employees and by customers must be considered.

## Product Quality Conformance

Products and services for our customers will conform to all requirements. Products will meet performance specifications. Services will be complete, meet described requirements, and will be in a format appropriate for the customer's use. If a specification cannot be met in full, the customer will be advised and a new specification will be negotiated.

## Product and Process Quality Improvement

All processes, manufacturing, manufacturing planning, customer service, product design and design of manufacturing processes shall utilize Total Quality Management concepts including Statistical Process Control techniques and designed experiments to ensure continual improvement of products and services.

## Employee Responsibility

Each employee is responsible for performing their work correctly and completely. This responsibility for quality performance applies to all design work, development work, manufacturing work and to all supporting work. It applies to all employee levels. It cannot be abandoned or delegated. No one else can take responsibility.

## IMP's Commitment of Support

IMP will provide the tools, the training, and the time necessary for employees to meet their responsibilities.

## Employee Participation

IM P encourages all employees to take part in the open discussion, analysis and resolution of problems through participation in quality and productivity teams or through personal suggestions.

# DET NORSKE VERITAS QUALITY SYSTEM CERTIFICATE 

Certificate No. 96-HOU-AQ-8474
This is to certify that the Quality System
of

## IMP INC.

at
2890 North First Street, San Jose, CA 95134 USA
Has been found to conform to Quality Standard:
ISO 9001, 1994
This Certificate is valid for the following products/service ranges:
DESIGN AND MANUFACTURE OF ANALOG AND MIXED-SIGNAL INTEGRATED CIRCUITS AND WAFER FABRICATION SERVICES

Mace and date:
Houston, Texas; 01 November 1996
for the Accredited Unit:
Det Noenke Veritaa Certification, Inc.
Houston, Texia, USA
DNV Management System Certification



Aceredited by the RwA.

This certificate is valid until:
08 August 1999

Initial Certification Date:


IMP offers higher performance, lower-power microprocessor supervisors that are pin compatible with devices from Dallas Semiconductor and Maxim Integrated Products. For the latest information visit www.impweb.com or send specific requests to info@impinc.com.

## P Supervisor Products: Low Power Alternatives to Maxim

| Part <br> Number | Threshold <br> Voltage (V) | Backup Battery Switch | Watchdog <br> Timer | Power Fail Monitor | Manual <br> Reset | RESET <br> Polarity |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IMP690A | 4.65 | X | X | X |  | LOW |
| IMP692A | 4.40 | X | X | X |  | LOW |
| IMP705 | 4.65 |  | X | X | X | LOW |
| IMP706 | 4.40 |  | X | X | X | LOW |
| IMP707 | 4.65 |  |  | X | X | L \& H |
| IMP708 | 4.40 |  |  | X | X | L \& H |
| IMP802L | 4.65 | X | X | X |  | LOW |
| IMP802M | 4.40 | X | X | X |  | LOW |
| IMP805L | 4.65 | X | X | X |  | HIGH |
| IMP809 | 2.63 to 4.63 |  |  |  |  | LOW |
| IMP810 | 2.63 to 4.63 |  |  |  |  | HIGH |
| IMP811 | 2.63 to 4.63 |  |  |  | X | LOW |
| IMP812 | 2.63 to 4.63 |  |  |  | X | HIGH |
| IMP813L | 4.65 |  | X | X | X | HIGH |

## Block Diagrams

IMP690A, IMP692A, IMP802L, IMP802M and IMP805L


IMP705, IMP706 and IMP813L


IMP707 and IMP708

$\mu$ P Supervisor Products: Low Power Alternatives to Dallas Semiconductor

| IMP Ordering Part Number | RESET <br> Voltage (V) | $\begin{gathered} \hline \text { RESET } \\ \text { Tolerance (\%) } \end{gathered}$ | RESET Time (ms) | RESET <br> Polarity | Push-Pull Output Stage | Open Drain Output | $\begin{aligned} & \hline \text { 8-Pin SO } \\ & \text { Package } \\ & \hline \end{aligned}$ | $\begin{gathered} \hline \text { TO-92 } \\ \text { Package } \\ \hline \end{gathered}$ | SOT-23 <br> Package | SOT-223 <br> Package |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IMP1810-5 | 4.620 | 5 | 150 | LOW | X |  |  | X |  |  |
| IMP1810-10 | 4.370 | 10 | 150 | LOW | X |  |  | X |  |  |
| IMP1810-15 | 4.120 | 15 | 150 | LOW | X |  |  | X |  |  |
| IMP1810R-5 | 4.620 | 5 | 150 | LOW | X |  |  |  | X |  |
| IMP1810R-10 | 4.370 | 10 | 150 | LOW | X |  |  |  | X |  |
| IMP1810R-15 | 4.120 | 15 | 150 | LOW | X |  |  |  | X |  |
| IMP1811-5 | 4.620 | 5 | 150 | LOW |  | X |  | X |  |  |
| IMP1811-10 | 4.350 | 10 | 150 | LOW |  | X |  | X |  |  |
| IMP1811-15 | 4.130 | 15 | 150 | LOW |  | X |  | X |  |  |
| IMP1811R-5 | 4.620 | 5 | 150 | LOW |  | X |  |  | X |  |
| IMP1811R-10 | 4.350 | 10 | 150 | LOW |  | X |  |  | X |  |
| IMP1811R-15 | 4.130 | 15 | 150 | LOW |  | X |  |  | X |  |
| IMP1812-5 | 4.620 | 5 | 150 | HIGH | X |  |  | X |  |  |
| IMP1812-10 | 4.350 | 10 | 150 | HIGH | X |  |  | X |  |  |
| IMP1812-15 | 4.130 | 15 | 150 | HIGH | X |  |  | X |  |  |
| IMP1812R-5 | 4.620 | 5 | 150 | HIGH | X |  |  |  | X |  |
| IMP1812R-10 | 4.350 | 10 | 150 | HIGH | X |  |  |  | X |  |
| IMP1812R-15 | 4.130 | 15 | 150 | HIGH | X |  |  |  | X |  |
| IMP1815-5 | 3.060 | 5 | 150 | LOW | X |  |  | X |  |  |
| IMP1815-10 | 2.880 | 10 | 150 | LOW | X |  |  | X |  |  |
| IMP1815-20 | 2.550 | 20 | 150 | LOW | X |  |  | X |  |  |
| IMP1815R-5 | 3.060 | 5 | 150 | LOW | X |  |  |  | X |  |
| IMP1815R-10 | 2.880 | 10 | 150 | LOW | X |  |  |  | X |  |
| IMP1815R-20 | 2.550 | 20 | 150 | LOW | X |  |  |  | X |  |
| IMP1816-5 | 3.060 | 5 | 150 | LOW |  | X |  | X |  |  |
| IMP1816-10 | 2.880 | 10 | 150 | LOW |  | X |  | X |  |  |
| IMP1816-20 | 2.550 | 20 | 150 | LOW |  | X |  | X |  |  |
| IMP1816R-5 | 3.060 | 5 | 150 | LOW |  | X |  |  | X |  |
| IMP1816R-10 | 2.880 | 10 | 150 | LOW |  | X |  |  | X |  |
| IMP1816R-20 | 2.550 | 20 | 150 | LOW |  | X |  |  | X |  |
| IMP1817-5 | 3.060 | 5 | 150 | HIGH | X |  |  | X |  |  |
| IMP1817-10 | 2.880 | 10 | 150 | HIGH | X |  |  | X |  |  |
| IMP1817-20 | 2.550 | 20 | 150 | HIGH | X |  |  | X |  |  |
| IMP1817R-5 | 3.060 | 5 | 150 | HIGH | X |  |  |  | X |  |
| IMP1817R-10 | 2.880 | 10 | 150 | HIGH | X |  |  |  | X |  |
| IMP1817R-20 | 2.550 | 20 | 150 | HIGH | X |  |  |  | X |  |
| IMP1233D-5 | 4.625 | 5 | 350 | LOW |  | X |  | X |  |  |
| IMP1233D-10 | 4.375 | 10 | 350 | LOW |  | X |  | X |  |  |
| IMP1233D-15 | 4.125 | 15 | 350 | LOW |  | X |  | X |  |  |
| IMP1233DZ-5 | 4.625 | 5 | 350 | LOW |  | X |  |  |  | X |
| IMP1233DZ-10 | 4.375 | 10 | 350 | LOW |  | X |  |  |  | X |
| IMP1233DZ-15 | 4.125 | 15 | 350 | LOW |  | X |  |  |  | X |
| IMP1233M-55 | 4.625 | 5 | 350 | LOW |  | X |  | X |  |  |
| IMP1233M-5 | 4.375 | 10 | 350 | LOW |  | X |  | X |  |  |
| IMP1233M-3 | 2.720 | 15 | 350 | LOW |  | X |  | X |  |  |
| IMP1233MS-55 | 4.625 | 5 | 350 | LOW |  | X | X |  |  |  |
| IMP1233MS-5 | 4.375 | 10 | 350 | LOW |  | X | X |  |  |  |
| IMP1233MS-3 | 2.720 | 15 | 350 | LOW |  | X | X |  |  |  |

## USB Power Switches

IMP offers a full complement of Universal Serial Bus (USB) power switches that are higher-performance equivalents to devices from Micrel.

| Part Number | \# of Switches | "ON" Resistance ( $\mathbf{m} \boldsymbol{\Omega}$ ) | Enable Polarity |
| :---: | :---: | :---: | :---: |
| IMP2525-1 | 1 | 140 | HIGH |
| IMP2525-2 | 1 | 140 | LOW |
| IMP2525A-1 <br> Low ON Resistance | 1 | 70 | HIGH |
| IMP2525A-2 <br> Low ON Resistance | 1 | 70 | LOW |
| IMP2526-1 | 2 | 140 | HIGH |
| IMP2526-2 | 2 | 140 | LOW |
| IMP2527-1 | 4 | 200 | HIGH |
| IMP2527-2 | 4 | 200 | LOW |

## EL Lamp Applications

- Pagers
- Caller ID
- Appliances
- Telephones

Thermostats
Weight Scales

- Cellular Phones
- Digital Compasses
- HPCs (Handheld PCs)
- Temperature Monitors
- Automotive Dashboards
- GPS Handheld Receivers
- PDAs (Personal Digital Assistants)
- Watches and Alarm Clocks
- Test and Medical Equipment


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## Sample Requests and New Product Updates

Name:

## Position:

Company:
Address:

| City: | State/Province: |
| :--- | :--- |
| Zip Code: | Country: |
| Phone \#: | Fax \#: |
| Email Address: |  |



## Fold

ISO 9001 Registered

## MS-200

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[^0]:    All devices are available in die form

